144MHz FM TRANSCEIVER

IC-25A/E

MAINTENANCE MANUAL

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SPECIFICATIONS

GENERAL

Numbers of semiconductors

Transistors

48

FETs

5

ICs

21 (IC-25A: 20)

Diodes

89 (IC-25A: 91)

Frequency coverage

144.000 ~ 145.995 MHz (IC-25A: 143.800 ~ 148.195 MHz)

Frequency resolution

5 kHz/25 kHz steps (IC-25A: 5 kHz/15 kHz steps)

Frequency control

Microcomputer based 5 kHz step Digital PLL synthesizer

Independent Dual VFO Capability.

Frequency stability

Within ± 1.5 kHz

Frequency channels Usable conditions

: 5 channels with any inband frequency programmable

Temperature:

 -10° C $\sim 60^{\circ}$ C (14° F $\sim 140^{\circ}$ F)

Operational time:

continuous

Antenna impedance

50 ohms unbalanced

Power supply requirement Current drain (at 13.8V DC) 13.8V DC ± 15% (negative ground) 6A Max.

Transmitting

LOW

HIGH

(25W) Approx. 4.8A

(1W)

Approx. 1.3A

Receiving

At max audio output

Approx. 0.6A

Squelched

Approx. 0.4A

50 mm (H) x 140 mm (W) x 177 mm (D)

Weight

Approx. 1.5 kg

Dimensions

TRANSMITTER

25W (HIGH), 1W (LOW)

Output power **Emission mode**

16F3

Modulation system

Variable reactance frequency modulation

Max. frequency deivation

: 5 kHz

Spurious emission

: More than 60 dB below carrier

Microphone

1.3 K ohm dynamic microphone with built-in preamplifier and

push-to-talk switch

Operating mode

Simplex, Duplex (Any inband frequency separation programmable)

Tone burst

: 1750 Hz ± 0.1 Hz (IC-25A: Not installed)

RECEIVER

Receiving system

Double-conversion superheterodyne

Modulation acceptance

16F₃

Intermediate frequency

1st:

: 2nd:

16.9 MHz 455 kHz

Sensitivity

More than 30 dB S+N+D/N+D at 1μ V

Less than 0.6µV for 20 dB Noise quieting

Squelch sensitivity

: Less than 0.4µV

Spurious response rejection ratio : More than 60 dB

Selectivity

: More than ±7.5 kHz at -6 dB point

Less than ±15 kHz at -60 dB point

Audio output power

: More than 2.0W

Audio output impedance

: 4 ~ 8 ohms

CONTROLS AND THEIR FUNCTIONS

FRONT PANEL

(1) VFO SWITCH

For selection of either VFO "A" or "B" for tuning.

In the "A" position, the VFO indicator illuminates and the frequency is changed in 5-kHz steps when the tuning control knob is turned. The frequency can be entered into any of the memory channels.

In the "B" position, the frequency is changed in 25-kHz steps (IC-25A: 15-kHz steps) when the tuning control knob is turned.

(2) VFO INDICATOR

This LED illuminates when VFO "A" is selected.

(3) PRIORITY INDICATOR

This LED illuminates when the priority function is switched ON.

(4) PRIORITY BUTTON

This switch is used to switch the priority function ON and OFF.

During a QSO or VFO reception, a calling channel or other desired channel can be monitored (whether the channel is in use or not) by setting the memory channel switch to that channel which has been memorized in a memory channel, and then press the priority button. The receiving frequency will thereafter automatically change to that channel for a moment every five seconds.

(5) FREQUENCY DISPLAY

The displayed frequency is the carrier frequency. The three large 7-segment LED's display the digits between 1 MHz and 10 kHz, and for 5 kHz of the operating frequency.

(6) RECEIVE INDICATOR

This LED illuminates when, during the RECEIVE mode, the squelch is opened.

(7) TRANSMIT INDICATOR

This LED illuminates in the TRANSMIT mode.

(8) VFO SCAN-FUNCTION SWITCH

This switch is used to select either full scan or program scan.

(9) S/RF INDICATOR

The seven in-line LED's indicate the S-unit and the RF output level. The digits of the S-meter represent S1 through S9 and 20 and 60 dB over S9. The RF output level meter functions only as a relative output meter; it does not indicate the wattage. These functions are automatically switched when T/R is switched.

(10) SIMPLEX/DUPLEX SWITCH

This switch is used to select either SIMPLEX operation or DUPLEX (repeater) operation.

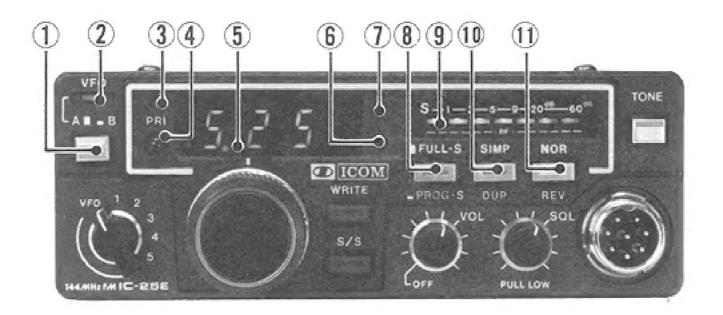
(11) DUPLEX-MODE SWITCH

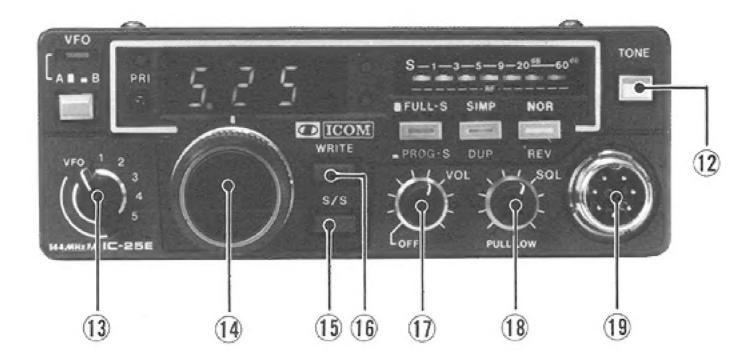
The switch is used to select the relationship of the reception frequency and the transmission frequency in the duplex mode.

In the "NOR" position, the transmission frequency is set to 600 kHz (This can be changed to any in-band frequency.) below the reception frequency. (IC-25A: The transmission frequency can be set to either above or below the reception by using the OFFSET switch.)

In the "REV" position, the reverse of the above is true for the transmission frequency and the reception frequency.

These operations can be used with the VFO or any of the memory channels.





(12) TONE CALL SWITCH (IC-25E only)

Most repeaters require a 1750-Hz tone burst for initial access. Pressing this switch for the required period for a repeater switches the unit to the transmit mode, and the tone-burst generator is activated, thus giving access to the repeater.

(12) OFFSET SWITCH (IC-25A only)

In the "+" position, the transmission frequency is set to 600 kHz (This can be changed to any in-band frequency.) above the reception frequency (when the DUPLEX MODE switch is in the "NOR" position).

In the "-" position, the transmission frequency is set 600 kHz below the reception frequency. In addition, the offset frequency can be changed to any in-band frequency (when the DUPLEX MODE switch is in the "NOR" position).

(13) VFO/MEMORY CHANNEL SWITCH

This switch is used to select either operation with the tuning control ("VFO" position) or operation with the memory channels (1 \sim 5).

(14) TUNING CONTROL KNOB

Turning this control clockwise increases the frequency, and turning it counterclockwise decreases the frequency. At the VFO "A" setting, the frequency is changed in 5-kHz steps, and at the VFO "B" setting, it is changed in 25-kHz (IC-25A: 15-kHz) steps. When this control is turned to a frequency exceeding 145.995 MHz (IC-25A: 148.195-MHz), the frequency will automatically revert to 144.000 MHz (IC-25A: 143.800 MHz). In the same way, when this control is turned to a frequency below 144.000 MHz (143.800-MHz), the frequency will automatically revert to 145.995 MHz (148.195 MHz).

(15) SCAN START/STOP BUTTON

This button is used to start and stop any of the scan functions. When it is pressed once again to restart the scan, the scan will start from the memory channel or frequency where it was stopped.

(16) MEMORY/VFO WRITE BUTTON

When this button is pressed, the VFO A frequency is "written" into a memory channel, or the frequency of one VFO is transferred to the other VFO.

(17) VOLUME CONTROL/POWER SWITCH

The power is OFF when this control is turned completely counterclockwise. The power can be switched ON by turning this control clockwise until a "click" is heard. The audio level is increased as the control is turned further clockwise.

(18) SQUELCH CONTROL/RF POWER SWITCH

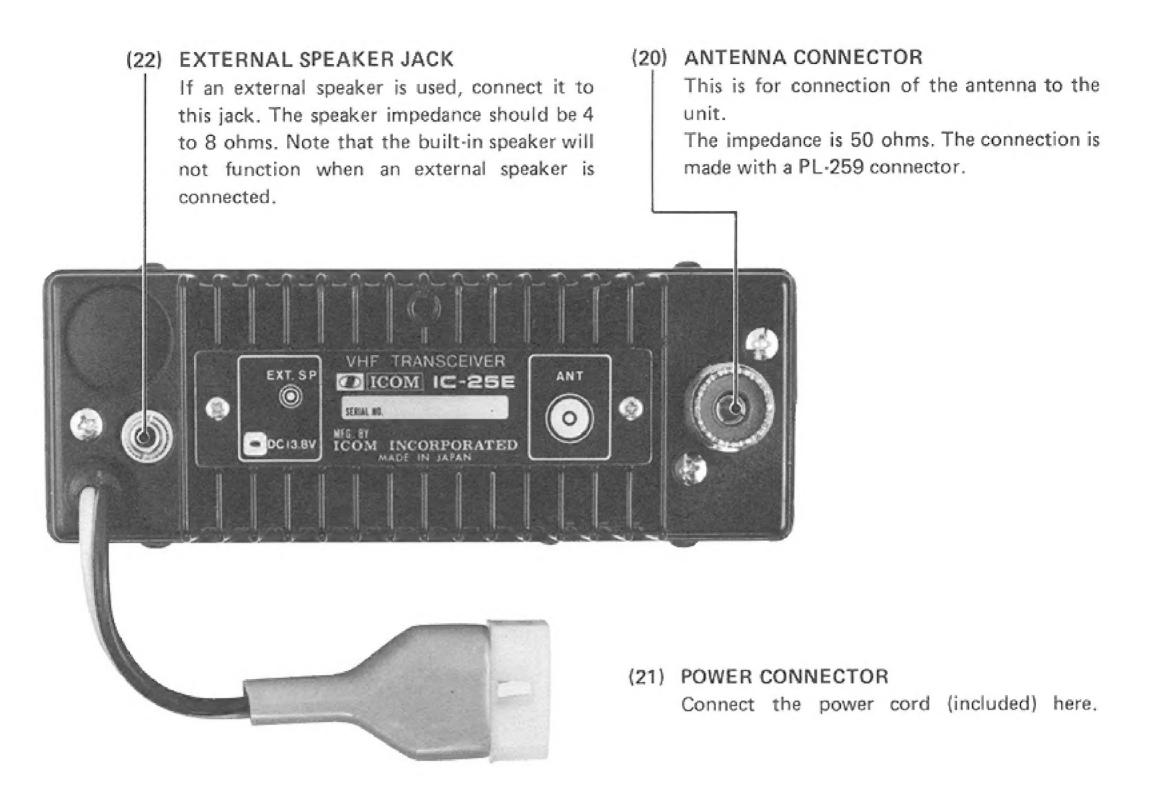
The squelch threshold is increased by turning this control clockwise. Turn it completely counterclockwise to turn the squelch function OFF.

When this knob is pulled outward, the RF output power is reduced to 1 watt. When it is pressed inward to the normal position, the RF output power is returned to 25 watts.

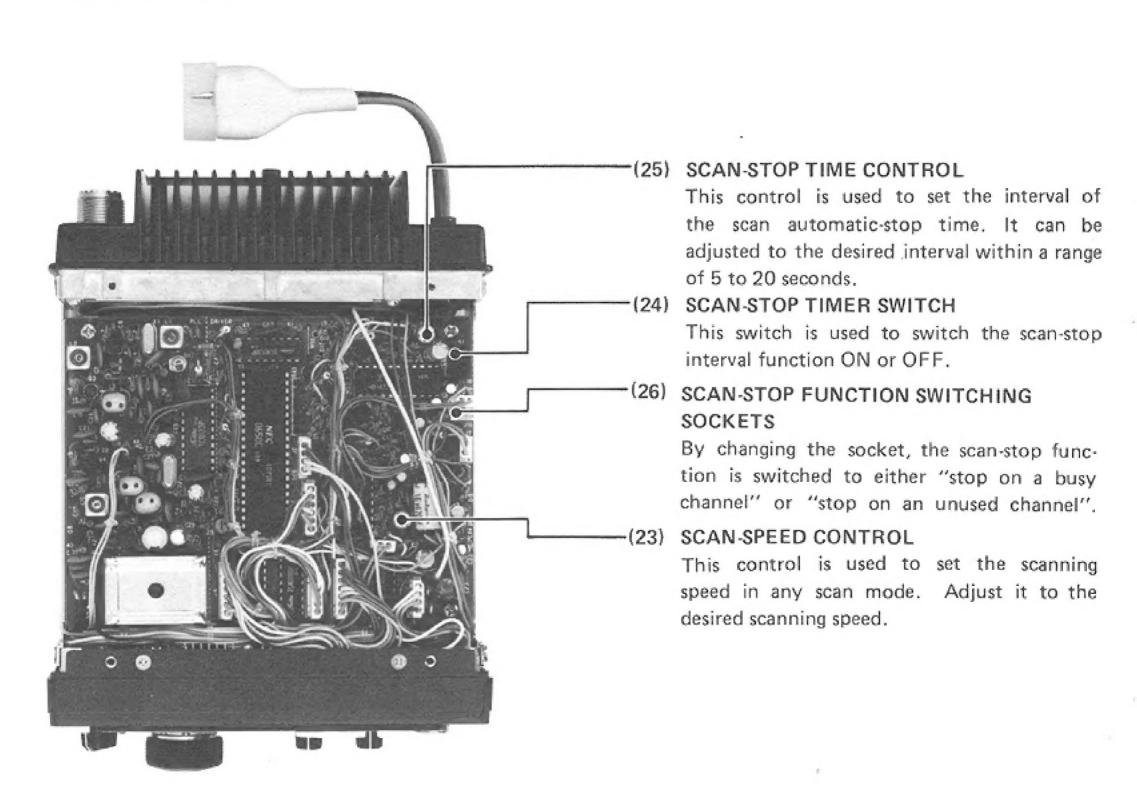
19) MIC CONNECTOR

Connect the microphone (included) to this connector.

(If you want to use a different microphone, refer to the illustration on page 5.)

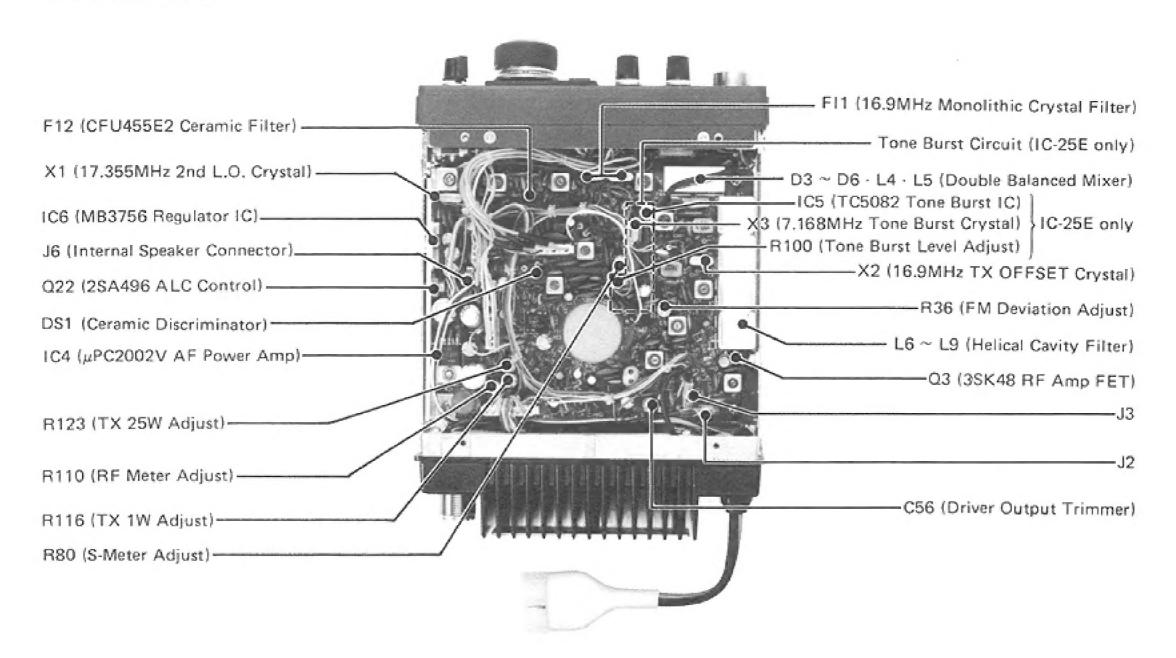


UNDER THE TOP COVER

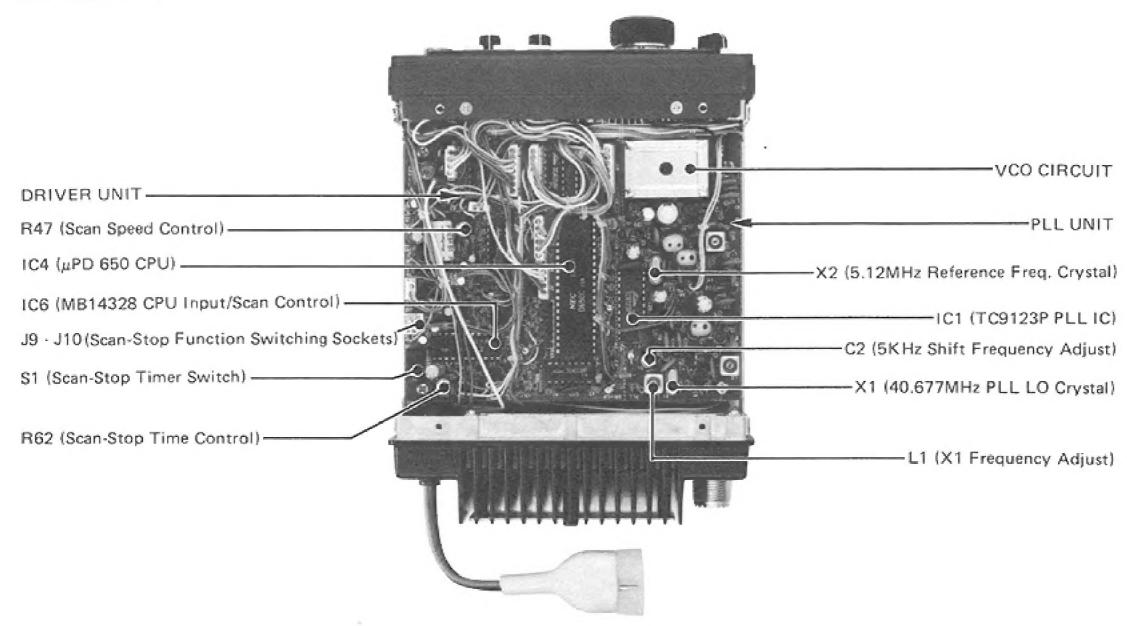


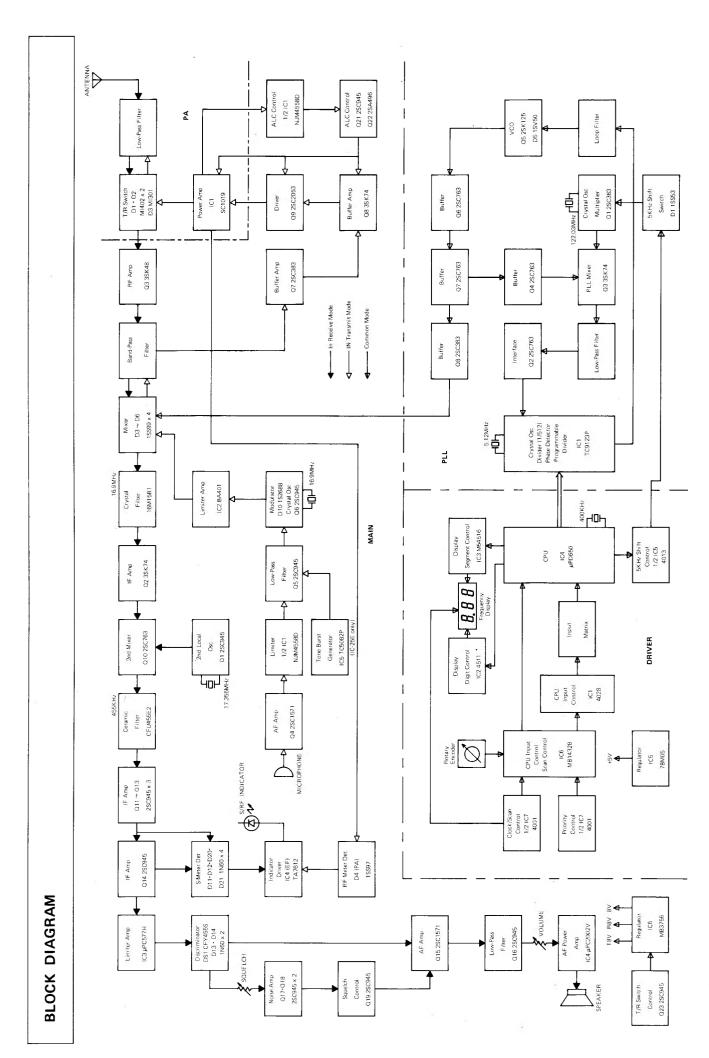
INSIDE VIEWS

MAIN UNIT SIDE



PLL UNIT SIDE





OPERATION

Model IC-25A/E is a 144-MHz FM transceiver composed of the main unit, PLL unit, driver unit, PA unit, etc.

Its fundamental circuitry is the double-conversion superheterodyne type, with a 1st IF of 16.9 MHz and a 2nd IF of 455 kHz. Frequency control is made by the microcomputer (incorporated in the driver unit) and the PLL unit which it controls.

Receiver

(1) PA unit

The PA unit is composed of the low-pass filter, T/R switching, power amplifier circuitry, etc.

The signal input from the antenna passes through the low-pass filter (consisting of L3 \sim L5 and C13 \sim C17), and is then led to the main unit RF circuitry after passing through a constant-K π -type filter (composed of C9, L2 and C5 by D3, which is ON only during reception).

(2) Main unit (RF circuitry)

The RF unit is composed of RF amplifier, a band-pass filter, mixer circuitry, etc.

The receiving signal, which has passed from the PA unit through J2 is amplified by approximately 18 dB by MOS FET (3SK48) Q3, which features a low noise figure and intermodulation characteristics.

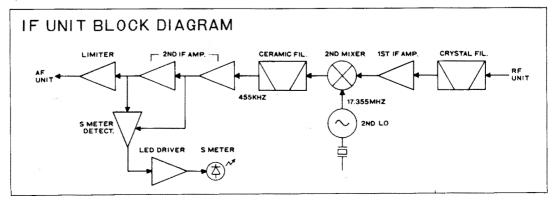
Unwanted signals are removed from the RF-amplified signal by the helical cavity band-pass filter (L6 \sim L9), after which the signal is input to the mixer circuitry of the next stage. This mixer circuitry features a wide dynamic range, and employs a double balanced mixer (DBM) composed of Schottky diodes (D3 \sim D6) which feature superb isolation between each port. The signal is mixed at this DBM with the 127 \sim 9-MHz signal input from the PLL unit, and converted to the first IF.

(3) Main unit (IF circuitry)

The IF unit is composed of a filter, the 1st IF amplifier, 2nd mixer, 2nd IF amplifiers circuitry, etc.

The crystal filter, which features superb selectivity, takes out only the wanted signals from the receiving signal of the RF unit after the signal has been frequency-converted by the DBM, and then, after amplification of approximately 20 dB by Q2, it is input to Q10 of the 2nd mixer circuitry.

The signal is then, at Q10, mixed with the 17.355-MHz signal oscillated by Q1, after which the 455-kHz 2nd IF signal is output. This signal output from the 2nd mixer passes through ceramic filter FI2, is IF-amplified by amplifier circuitry composed of Q11 \sim Q13, after which it passes through IC3 limiteramplifier.

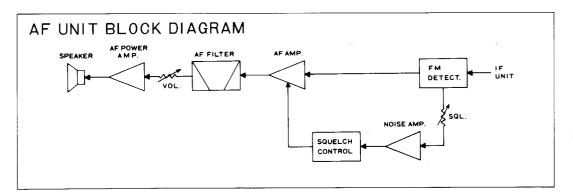


(4) Main unit (AF-amplification circuitry)

The AF-amplification unit is composed of circuitry for FM detection, AF amplification, AF power amplification, etc.

The signal input by limiter-amplifier IC3 is detected by the frequency-discrimination circuitry (composed of DS-1, D13, D14, R82 and R83), and is then de-emphasized at the integration circuitry (R85 and C78).

This signal is AF-amplified by Q15, passes through the low-pass filter formed by Q16, where unnecessary components are removed, and then, via the volume control, is amplified to the level which will drive the speaker by IC4 (for AF amplification), so that the speaker is thereby driven.



(5) S-meter circuitry

In the S-meter circuitry, a weak signal passes from the L21 center tap and through C66 where it is taken out and rectified by D11 and D12. Because strong signals are detected by D20 and D21, the circuit has a wide dynamic range.

(6) Squelch circuitry

Signals detected by discriminator DS-1 pass through the squelch control, the noise component only is amplified by Q17, and then Q19 is switched as a result of D16 and D17 detection, and the Q15 (AF amplifier) is switched.

2. Transmitter

(1) Microphone amplification and modulation unit

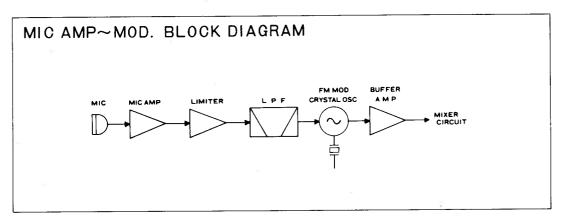
This unit is composed of microphone amplifier, limiter-amplifier, FM modulation circuitry, etc. Modulation is applied when 16.9 MHz is oscillated by X2 and Q6, and the output from Q5 is applied to varactor diode D10. Residual AM is removed from the modulated signal by IC2 (limiter amp.) In the next stage, after which the signal is input to the double balanced mixer (used also for reception) composed of Schottky diodes.

(2) Band-pass filter and Younger stage (YGR) unit

The signals mixed by the DBM pass through a band-pass filter (used also for reception) composed of L6 \sim L9, where nearby spurious signals are removed.

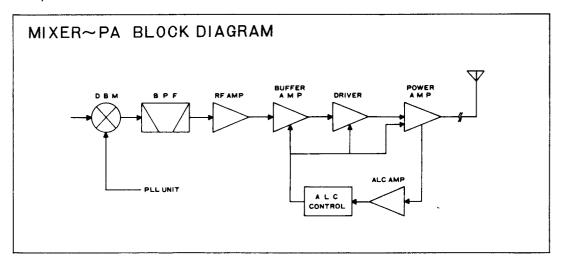
Signals which have passed through this filter are applied, via T/R switch D7, to Q7.

Next, they are amplified by the YGR unit (composed of Q7, Q8 and Q9), and are power-amplified to the PA module drive level (200mW).



(3) PA unit

Input signals from the YGR unit are amplified to approximately 25W by the PA module IC1 (SC1019), after which they pass, via D1 and D2 (ON during transmission), through the resonance circuitry (composed of C9 and L2). The low-pass filter (composed of L3 \sim L5 and C13 \sim C17), with Chebyshev characteristics, suppresses spurious signals by approximating the maximum attenuation points to the second and third harmonics, after which the transmission signal is supplied to the antenna's circuitry.



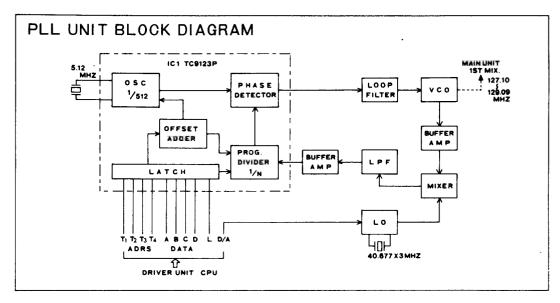
(4) ALC unit

The final-stage current of the PA section is detected as a result of the R1 voltage drop, and is input to IC1 (pin 5) of the main unit. In addition, HV line is applied to pin 6 of IC1, so that the operation voltage of the excitation amplifier (Q8 and Q9) and PA module IC1 is thereby controlled by IC1, Q21 and Q22.

(5) Power-supply unit

The power-supply circuitry IC6 is a voltage regulator IC (MB3756) to assure a stable voltage supply. An input of 13.8V is input from pin 2, and a regulated 8.2V is supplied to +8V (pin 1), R8V (pin 6) and T8V (pin 8) lines. Note that Q23 is, by stand-by muting, the IC6 control transistor.

3. PLL unit



The PLL unit of this model is mixed-down PLL circuitry controlled by the CPU of the driver unit. A frequency of 40.677 MHz is oscillated in the local-oscillator circuitry, and a frequency of 122.03 MHz (40.677 x 3) is obtained. The VCO is locked at each 10-kHz interval through the PLL frequency range of $127.10 \sim 129.09$ MHz.

(1) Local-oscillator circuitry

The oscillation frequency of the local-oscillator circuitry is obtained according to the following formula:

fo: VCO output frequency fm: local-oscillator frequency

N: programmable divider divisions (see table 1)

fi: reference frequency (10 kHz)

(1) Converted to the formula:

$$fm = fo - Nfi - (2)$$

For a carrier frequency of 144 MHz, the VCO output frequency fo:

$$fo = 144 (MHz) - 16.9 (MHz) = 127.1 (MHz)$$

* 16.9 MHz is the first IF frequency

(2) Which is expressed as:

fm = 127.1 (MHz) - 507 x 10 (kHz) = 127.1 (MHz) - 5.07 (MHz) = 122.03 (MHz)

By the above, then, the local-oscillator frequency becomes 122.03 MHz at 144 MHz. The frequency oscillated by X1 and Q1 is tripled and tuned by the tuning circuitry consisting of C8 and L2, with the result that X1 (VCO crystal) oscillates a frequency of 40.676 MHz $(122.03 \div 3)$.

Table 1

CPU output	Division N
400	507
401	508
499	606
500	607
501	608
:	
599	706
•	
700	807
į	į
800	907
	400 401

(2) PLL mixer circuitry

A high-gain, low-noise dual-gate MOS FET is employed in mixer Q3, thereby reducing local-oscillation leakage.

Only the difference component is taken out by low-pass filter (composed of L3, C13 and C14) in the next stage. This signal is then amplified by Q2 to the level at which the programmable divider is enabled, and is then input to IC1 (pin 12).

(3) Reference frequency oscillation division, phase comparison and programmable divider

IC1 is a multi-function IC which incorporates the above functions in one package.

The reference frequency of 10 kHz oscillates 5.12 MHz by the crystal attached at IC1 (pins 20 and 21), and by a frequency division of 1/512, an accurate reference signal of 10 kHz is obtained.

Digital phase comparison of this reference signal and the signal which has been mixed-down by the PLL mixer circuitry and frequency-divided by the frequency division N of the programmable divider is made by the phase comparator, and the result is output to pin 16.

As for the programmable divider input, the frequency data (BCD code) output from the CPU goes to the A \sim D terminals (pins 6 \sim 9), the digit-assigned data is prepared for T1 \sim T3 (pins 2 \sim 4), and the readout is performed, according to the timing of the load enable pulse (L terminal), in the order beginning from the most-significant digit.

Note that, because the counter offset of this IC is +107, the frequency and divider frequency division N become just as shown in table 1.

(4) Loop filter circuitry

The phase comparator output, taken from pin 16 of IC1, pulses in accordance with the phase difference, and, for that reason, the harmonic component and noise component are removed, together with DC conversion, as the signal passes through the lag-lead filter (composed of R26, R27, R28, C28, C29 and C31). When the output voltage is rapidly changed by a large value by jumping the frequency from the upper to lower band edges or vice versa, D3 or D4 is turned ON and the output signal does not pass through R26, and C29 is charged directly, with the result that response becomes quicker.

In addition, so that there will be response to the positive and negative pulses from the phase comparator, each is attached at reverse polarity. And, in constant operation, D3 and D4 are OFF, and the loop filter width is narrow, so that there is little influence by surrouding noise, etc.

(5) VCO and buffer-amplification circuitry

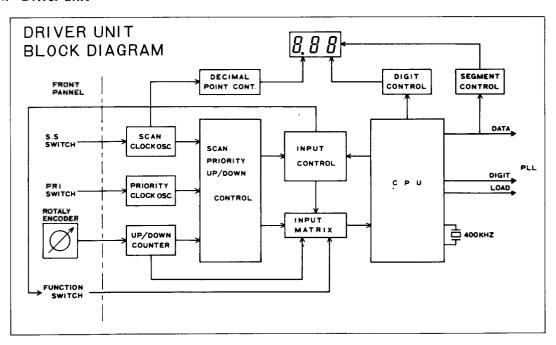
The voltage changed to DC by the loop filter is supplied to varactor diode D5 of VCO circuitry, and is subjected to control the oscillation frequency. This VCO is a Colpitts type of oscillation circuitry consisting of Q5 and a tuned circuit, and the signal is buffer-amplified by Q6 and Q7 in the next stage. Because the VCO output is used as the 1st local oscillator signal for transmission and reception, the impedance is matched with DBM, and the signal is amplified until the conversion loss of DBM is minimized by Q8.

In addition, the output taken from the center tap of L10 passes through isolation amplifier Q4, and is injected at the second gate of Q3 PLL mixer, thereby reducing leakage from the PLL local oscillator.

(6) Transmission muting circuitry

When the lock is unlocked, pin 18 of IC1 becomes ground level. By taking advantage of this, unwanted waves are prevented by stopping T8V of the main unit from being applied.

4. Driver unit



The driver unit, the CPU, as the main device, is composed of circuitry for CPU input control, input matrix, display, etc.

The up/down and clock signals obtained from the rotary encoder (directly coupled to the tuning control knob on the front panel) are sent in order to the CPU. These signals are handled at the CPU according to the program, and display data and frequency control data are sent to the frequency display and the PLL unit.

Main functions of logic unit IC's transistors and diodes

101	Conversion of BCD code to 10 digite: 0 output
IC1	Conversion of BCD code to 10 digits; Q output
IC2	For display; BCD code conversion to 7-segment code
IC3	Transistor array; display digit control
IC4	CPU
IC5	CPU reset
IC6	Input control; control of sensor data, mic up/down data, timer, priority, scan,
	etc.
IC7	2 input NOR x 4; unstable multi-vibrator for priority and scanning
IC8	INV x 6; control of busy/unused, VFO5, etc.
Q1 & Q2	Memory scan ← full/prog. scan switching
Q4	Power ON (CPU)
Q5	For memory read
Q6	For priority release at VFO
Q7	For priority start at memory ch.
Q8	For auto-stop during priority operation
Q9	For scan speed reduction during memory ch.
QI0	For scan clock and sensor input
Q12	For control of CK and UD (IC6 output)
Q13	For timer
Q14	For S power supply
Q15 & Q19	For decimal point flashing during scanning
Q16 & Q17	For 5V during VFO operation
D1 & D4	Memory ←→ full/prog. scan switching
D1, D3 & D6	For voltage drop prevention
D9 & D10	For offset frequency setting during initial operation
D14	For priority auto-stop
D24	For decimal point flashing during scanning
D27	1 count when scan stops
D32 & D33	For load reduction during initial operation
D34	For prevention of latch-up
D36	For band setting during initial operation
-	

Operation of logic unit IC6 (MB14025)

This IC is a custom LS type TTL, with ICOM's own unique circuitry in a 22-pin plastic package. Its main functions are the scan and input controls and interface for IC4 (CPU).

This IC controls each operation by the clock input to pin 14 (STM).

Pin	Name	Operation
1.	N.C.	Not used
2.	N.C.	Not used
3.	SUD	Sensor data up/down and scan input control
4.	SCK	Sensor data count and scan input control
5.	SCO	Output of S/S input (pin 8)
6.	MSL	For reset
7.	MCL	For reset
8.	S/S	S/S switch input
9.	RSW	Internal latch input
10.	SEL	Timer ON/OFF input
11.	GND	Ground
12.	BUSY	Used as ground
13.	SQL	Squelch input
14.	STM	Custom clock input
15.	ITI	Timer input
16.	ITO	Timer output
17.	RIC	Internal latch output
18.	RLD	For PRIO LED
19.	RCK	Control of up (pin 20) and CK (pin 21)
20.	UD	SUD output
21.	CK	SCK output
22.	+Vcc	Power supply (+4.5 \sim 5.5 V)

(1) Up/down, scan and external up/down control circuitry

The signals (data) generated by operation of the tuning control knob, the scan start/stop button, microphone scan, etc., pass through the pulse-generation circuitry (consisting of IC7, IC8, Q9, Q10 and Q11), the waveform-shaping circuitry, and the A/D conversion circuitry, and are then input to the input control TTL IC (IC6), developed by ICOM with its own unique program written in.

AT IC6, these input data are quickly and precisely fed to the CPU.

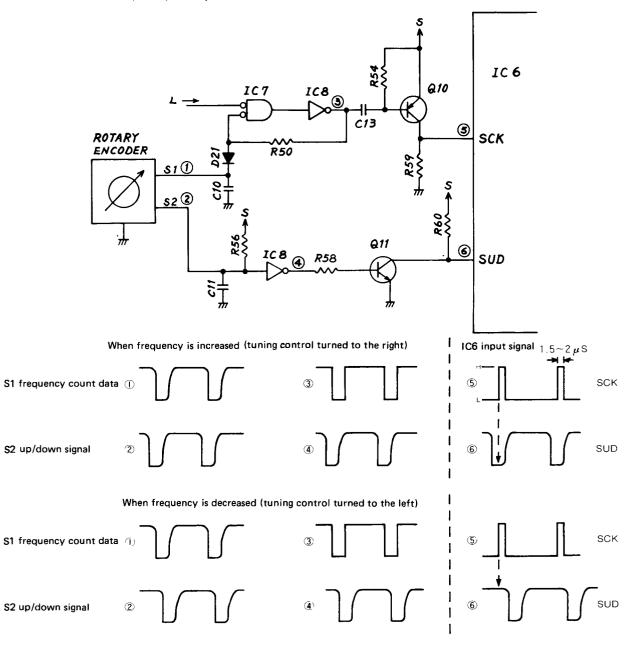
Frequency up/down circuitry

By turning the tuning control knob clockwise or counterclockwise, signal S1 (for counting the frequency), and signal S2 (up or down) which distinguishes the turning direction are generated at the rotary encoder. S1 and S2 have a 90° phase difference, so that S2 is advanced from S1 when the tuning control is turned to the right, and is delayed when it is turned to the left.

The S1 signal (1) is integrated by the rotary encoder's internal resistor (approx. 600Ω) and C10, and then, after removing contact chattering, it is formed into a perfectly square wave by the Schmitt circuitry (consisting of IC7 and IC8), is differentiated at (3) C13 and R54, and is input to the IC6 SCK with a pulse width of 1.5 to 2μ S. An addition (if the tuning control was turned to the right) or subtraction (if turned to the left) of an initialized 5 kHz or 25 kHz (IC-25A: 15 kHz) occurs for each pulse input at the SCK.

The S2 signal, which makes addition or subtraction, is integrated by the encoder's internal resistor and C11, and then, after chattering has been removed, the signal (4) is input at SUD (6) IC6.

When the S1 pulse is input to the SCK, the action will be an addition (up) if SUD (S2 pulse) is H level, or subtraction (down) if the pulse is L level.



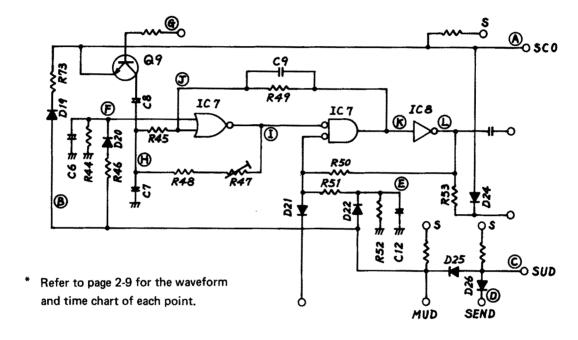
Microphone up/down and scan control circuitry

The voltage of SCO is reduced when the scan start/stop button is pressed, and because the voltages of (F) and (E) are reduced as a result, the multi-vibrator (IC7) oscillates. At this time, SUD becomes the H level according to the voltage across R73 and the junction voltage of D19 and D25, and the down-scan action is activated. Because the time-constants of C6 and R44 and of C12 and R52 differ, continuous scan starts after the first count of one.

As for memory scan, the scan speed becomes slower because C7 and C8 are in parallel. Note that R47 is the control for adjustment of scan speed.

Microphone up/down is accomplished in the same way. When it is up, only the junction voltage of D25 is applied to the SUD terminal, and, when it is down, a voltage divided by R60 and 470 Ω (H-level) is applied to the SUD.

D26 is a diode which stops the scan operation during transmission. Actually, microphone up-scanning only is possible during transmission.



Priority-control circuitry

When the power is turned on, IC6's RSW, RIC and RLD become H level. When the VFO/MEMORY CHANNEL switch is set to the VFO position, the Q7 base becomes L level, causing Q7 to switch OFF, so that RSW will not become L level even if the priority button is pressed.

When the VFO/MEMORY CHANNEL switch is set to a memory channel (1 \sim 5), Q7 is switched ON, and when the priority button is pressed once, an L pulse enters RSW, and the priority circuitry becomes ON. In this condition, both RIC and RLD become L level.

When RIC becomes L level, the IC7 (A) and (B) oscillation circuitry operates.

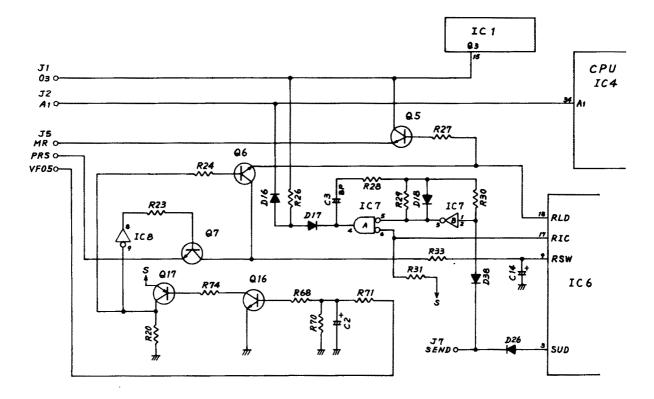
When RLD is L level, Q5 is switched OFF, and, unless pin 4 of IC7 becomes H level, the condition is of VFO A or B.

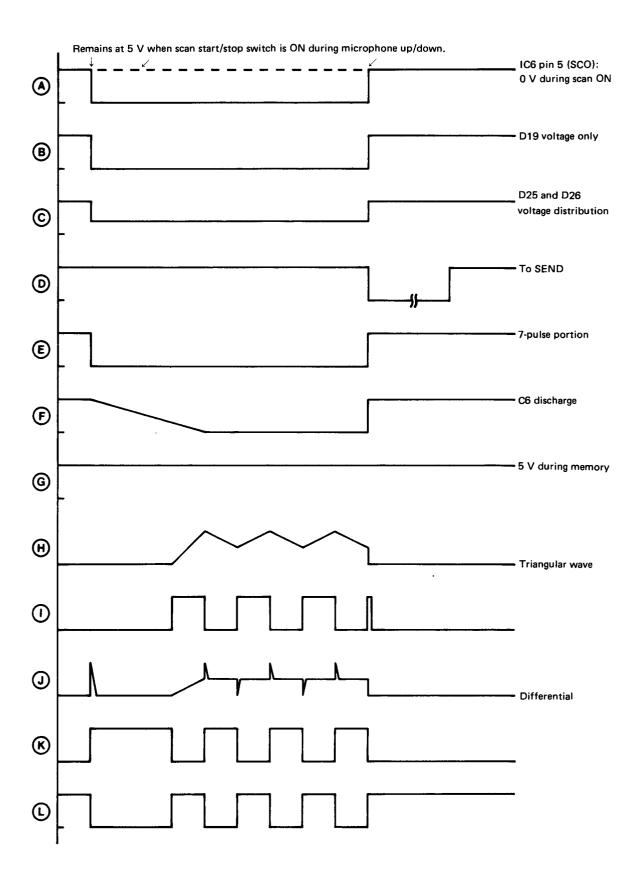
When pin 4 of IC7 becomes H level, the flow in the input matrix (consisting of R26 and D16) becomes $Q_3 \rightarrow A_1$, and the designated memory read-out occurs.

When transmission occurs during the priority condition, SUD is changed to L level by D26, and, as a result, RIC becomes H level. As a consequence, the oscillation of IC7 (A) and (B) stops, and pin 4 of IC7 becomes L level. However, because the level of RLD is maintained even during a condition of oscillation, the priority LED remains illuminated.

When the unit changes back to reception, RIC becomes L level, and IC7 (A) and (B) again resume their oscillation.

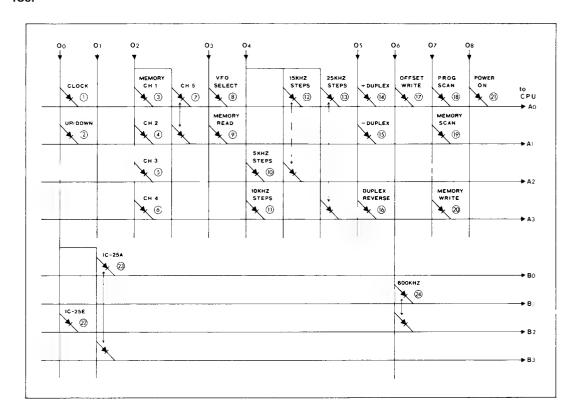
When, during the priority condition, the VFO/MEMORY CHANNEL switch is switched from the memory channel to VFO, the Q6 base changes from L level to H level, so that RSW is changed to L level by Q6, and the priority condition is released.





(2) CPU input-control and matrix circuitry

In order for the CPU to catch the various input data activated by the switches and buttons on the front panel, others for up/down and scan, input matrix circuitry and IC1 for input control are incorporated. Input data pass through this circuitry and are input to the CPU. When, however, there are many input data, the 4-bit configuration limits discrimination to 16, so that data discrimination is performed by a time-sharing operation by a timing pulse ($QO \sim QB$) sent out from IC1 as data pass through input-control IC6.



(1) O0 → A0 (Clock)

This flow occurs when the frequency is moved up or down with each pulse input by turning the tuning control knob or scanning.

(2) $O0 \rightarrow A1 (Up/Down)$

This flow occurs when the frequency is moved up. When the frequency is moved down, the O0 signal is not fed to A1.

(3) O2 → A0 (Memory Channel 1)

This flow occurs when the VFO/MEMORY CHANNEL Switch is set at "1".

(4) O2 → A1 (Memory Channel 2)

This flow occurs when the VFO/MEMORY CHANNEL Switch is set at "2".

(5) O2 → A2 (Memory Channel 3)

This flow occurs when the VFO/MEMORY CHANNEL Switch is set at "3".

(6) O2 → A3 (Memory Channel 4)

This flow occurs when the VFO/MEMORY CHANNEL Switch is set at "4".

(7) O2 → A0, A1 (Memory Channel 5)

This flow occurs when the VFO/MEMORY CHANNEL Switch is set at "5".

(8) O3 → A0 (VFO Select)

This flow occurs when the VFO Switch is set in "B", and the unit operates at the frequency set by "B" VFO. When the VFO Switch is set in "A", the O3 signal is not fed to A0, and the unit operates at the frequency set by "A" VFO.

(9) O3 → A1 (Memory Read)

This flow occurs when the VFO/MEMORY CHANNEL Switch is set at a memory channel, and the set operates on the channel. This function is prior to the function (8).

(10) O4 → A2 (5KHz Steps)

This flow occurs when the VFO Switch is set at "A", and the operating frequency is moved up or down with 5KHz steps.

(11) O4 → A3 (10KHz Steps)

When this flow occurs, the operating frequency is moved up or down with 10KHz steps. (This mode is not used for IC-25A/E.)

(12) O4 → A0, A2 (15KHz Steps)

When this flow occurs, the operating frequency is moved up or down with 15KHz steps. (This mode is used for VFO B of IC-25A.)

(13) O4 → A0, A3 (25KHz Steps)

When this flow occurs, the operating frequency is moved up or down with 25KHz steps. (This mode is used for VFO B of IC-25E.)

(14) O5 → A0 (+Duplex)

This flow occurs when the OFFSET Switch is set at "+", and the transmit frequency becomes a frequency which is added the specified offset frequency to the receive frequency. (This mode is not used for IC-25E.)

(15) $O5 \rightarrow A1$ (—Duplex)

This flow occurs when the OFFSET Switch is set at "-", and the transmit frequency becomes a frequency which is subtracted the specified offset frequency from the receive frequency. (This mode is preset for IC-25E.)

(16) O5 → A3 (Duplex Reverse)

This flow occurs when the OFFSET Switch is set at "+", or "-" and the DUPLEX MODE Switch is in "REV" position, the receive frequency becomes a frequency which is added to, the specified offset frequency or subtracted it from the original receive frequency, and the transmit frequency becomes the original receive frequency.

NOTE: In the duplex operation, if an expecting transmit frequency becomes out of the band, this frequency will remain at the original receive frequency.

(17) O6 → A0 (Offset Write)

This flow occurs when the VFO/MEMORY CHANNEL is set at "VFO" position and the PRIORITY button is pushed, and the offset frequency can be reset by turning the tuning control knob.

(18) O7 → A0 (Programmed Scan)

When this flow occurs, and the S/S button is depressed, the operating frequency scans between frequencies written into the Memory Channels 1 and 2. (Memory Channel 2 should be written higher frequency than Memory Channel 1, if not the scan function does not actuate.)

When this flow does not occur and the S/S button is depressed, the operating frequency scans the entire band.

(19) O7 → A1 (Memory Scan)

When this flow occurs, the operating frequency scans on the memory channels and the VFO's A and B.

(20) O7 → A3 (Memory Write)

When a memory channel is selected and this flow occurs, a frequency of the VFO A is written into the selected memory channel.

When a VFO is selected and this flow occurs, a frequency of the other VFO is written into the selected VFO.

(21) $O8 \rightarrow A0$ (Power ON)

This flow occurs when the POWER Switch is turned ON. When this flow does not occur, the CPU is in stand by condition.

NOTE: The port B is used to initialize the CPU when the power is turned ON.

(22) $O0 \rightarrow B2$

When this flow occurs, the operating frequency range is selected between 144.000MHz and 145.995MHz. (This mode is used for IC-25E.)

(23) $O0 \rightarrow B0, B3$

When this flow occurs, the operating frequency range is selected between 143.800MHz and 148.195MHz. (This mode is used for IC-25A.)

(24) O6 → B1, B2

When this flow occurs, the offset frequency is preset for 600KHz.

(3) CPU and output

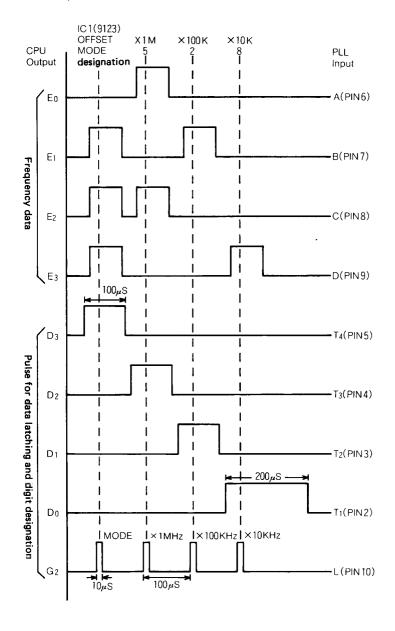
The CPU performs several operations depending on the programs written into it: initial setting, memorization, calculation, and output processing. The data input from terminals A0 \sim A3 and B0 \sim B3 are instantaneously processed according to the program. The data are then output from output terminals E0 \sim E3, F0 \sim F3 and D0 \sim D3, and the individual data are then sent on to frequency display and PLL unit. The data output from terminals F0 \sim F3 are the frequency data, and are displayed, via IC2 (for display drive), on the display. These data are also sent to IC1 (A \sim D terminals) of the PLL unit, there becoming the PLL frequency data. IC2 converts the CPU output data to character display segment (a \sim g) data, and transfers the data to the display. IC3, based upon the data output from F0 \sim F2 of the CPU, designates the display digits.

The output data from D0 \sim D3 are sent to IC1 of the PLL unit, and are there used to designate the frequency digits. Note that the following time chart should be used as reference concerning the relationship between the F0 \sim F3 and D0 \sim D3 output and the PLL.

CPU → PLL frequency data transfer time chart

- Frequency data are sent from E0 ~ E3 of the CPU to terminals A ~ D of IC1 (TC9123P) of the PLL.
- The digit signal to the PLL is output from D0 \sim D3 of the CPU, and is input at T1 \sim T4 of the PLL IC1.
- Load pulses are sent out one after another from G2 of the CPU, and the data are latched according to the timing of the pulses.

Example: FM 145,280MHz



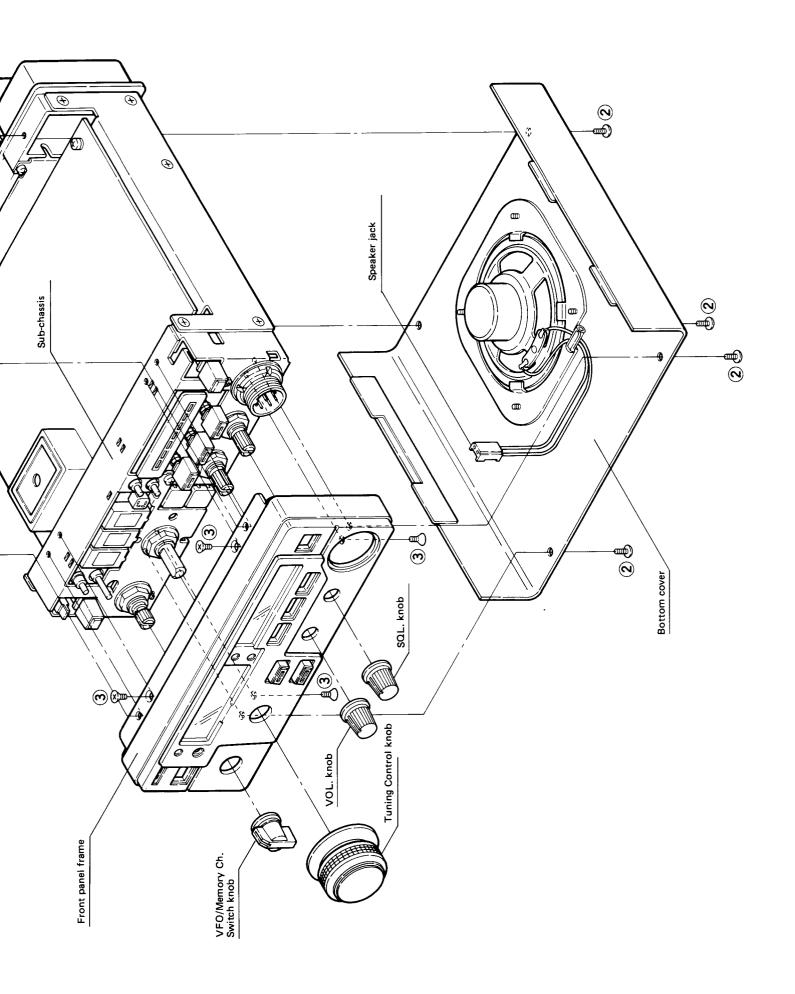
Logic unit IC's

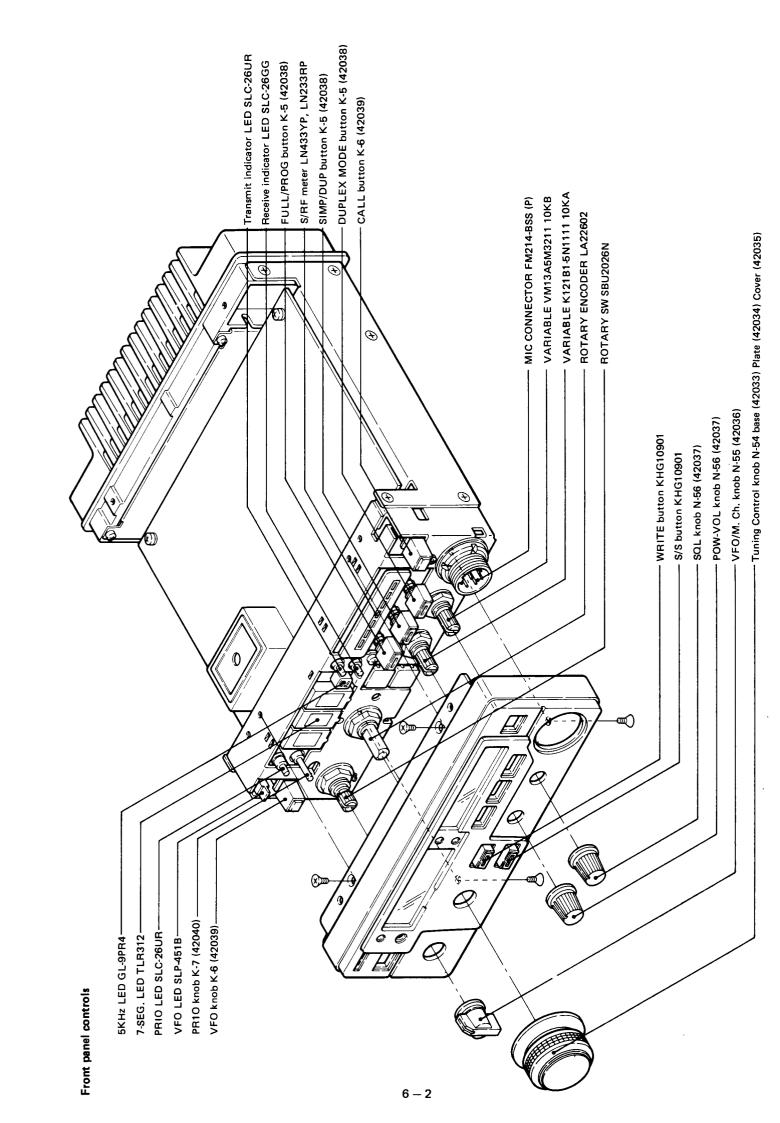
The IC's used in the logic unit are TTL (IC6 only) and C-MOS (except IC6).

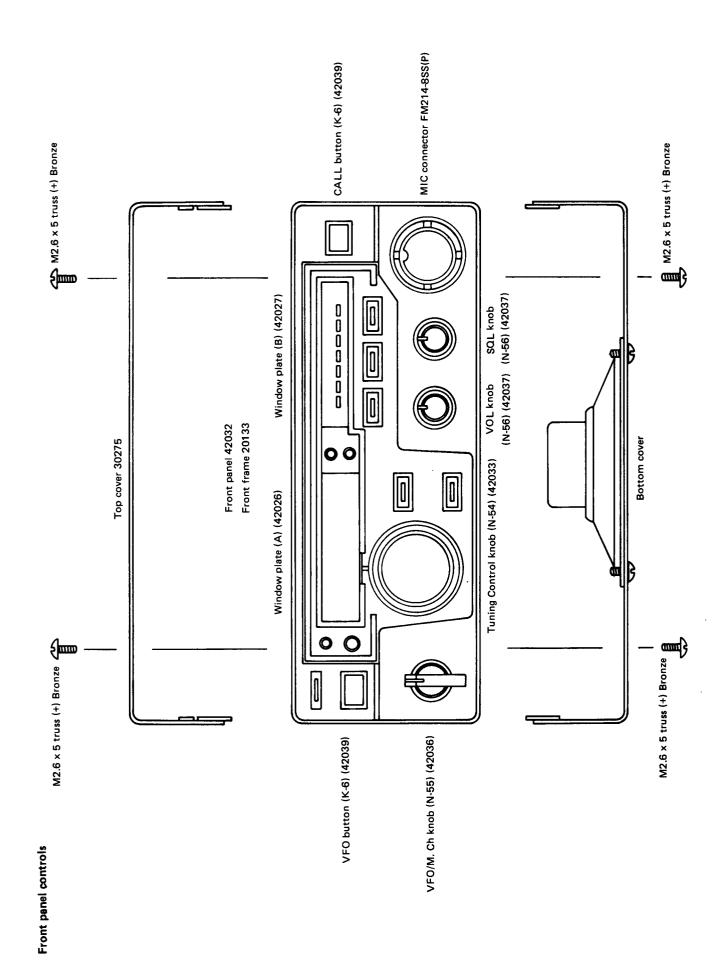
Because it is difficult to check with a tester each part which operates digitally because of the pulse signals, it is very important to know the operation timing of each part, and to know the H or L level operation points.

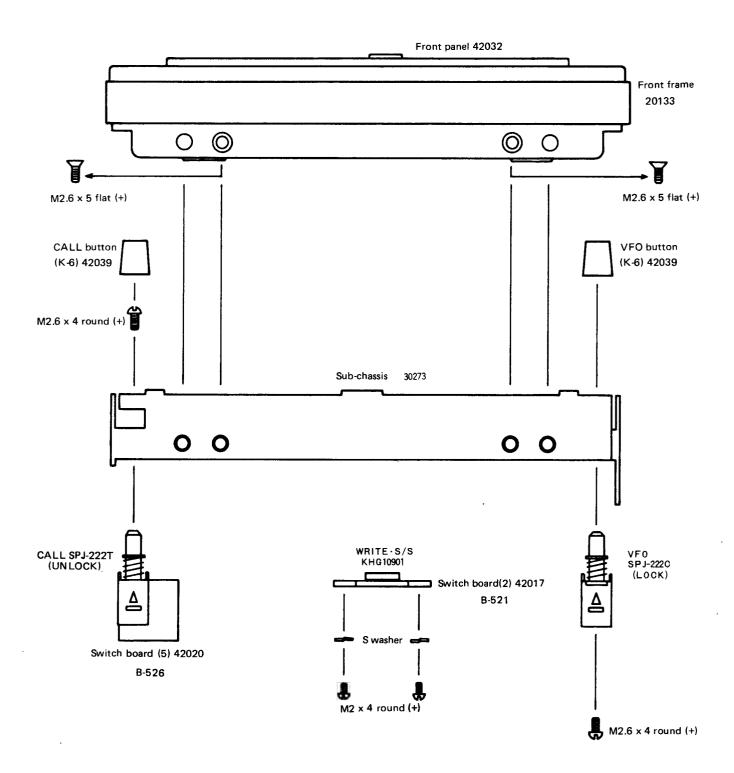
The IC threshold voltages are as follows:

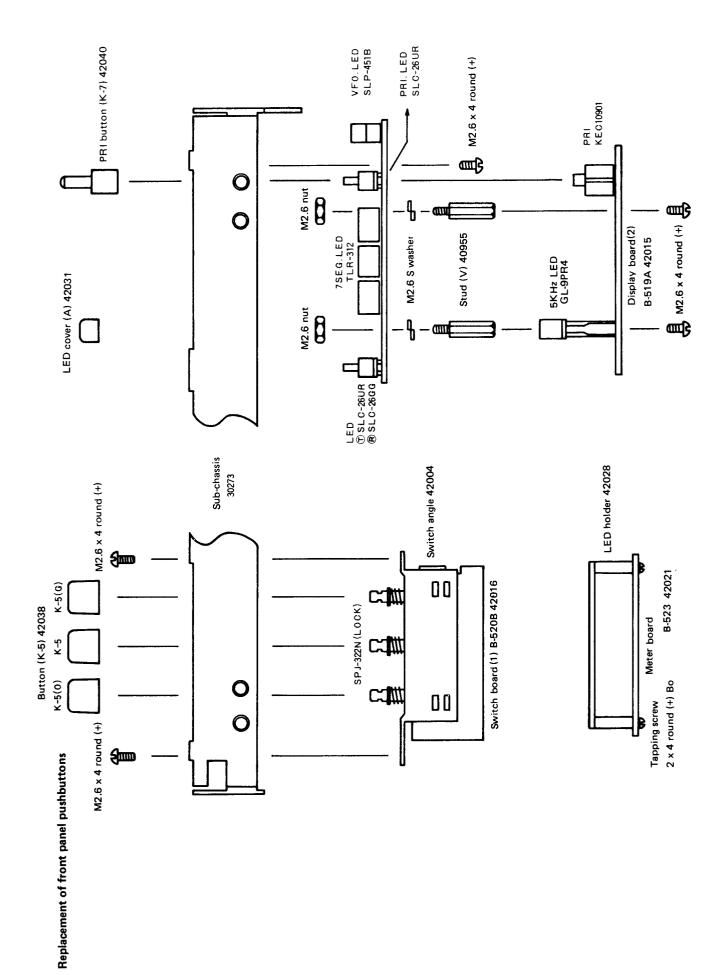
IC6 (TTL) Vcc: 5V Threshold: 1.5V 0.8V L level input (max.): H level input (min.): 2.0V Except IC6 (C-MOS) VDD: 5V Threshold: 2.5V 1.5V L level input (max.): H level input (min.): 3.5V



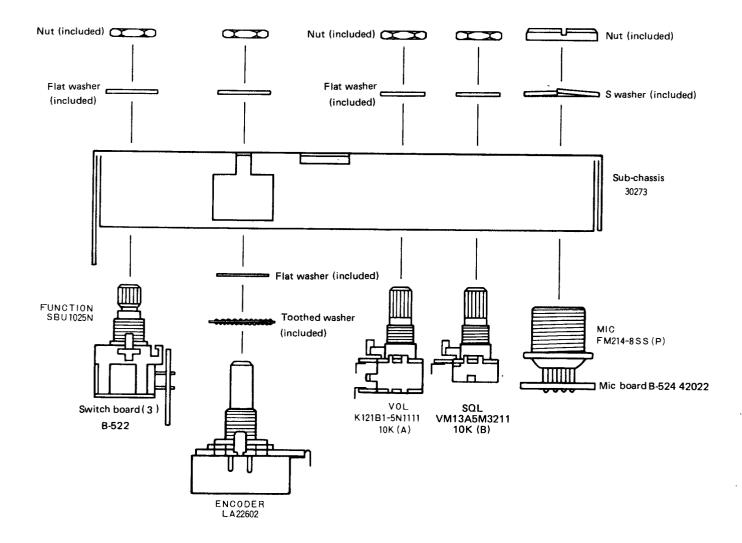


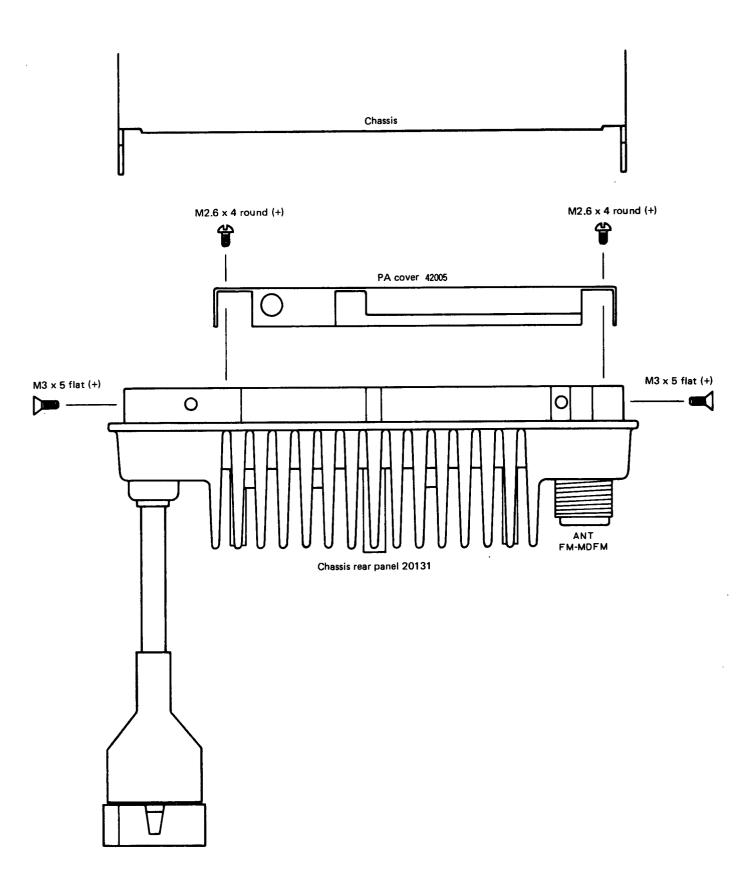




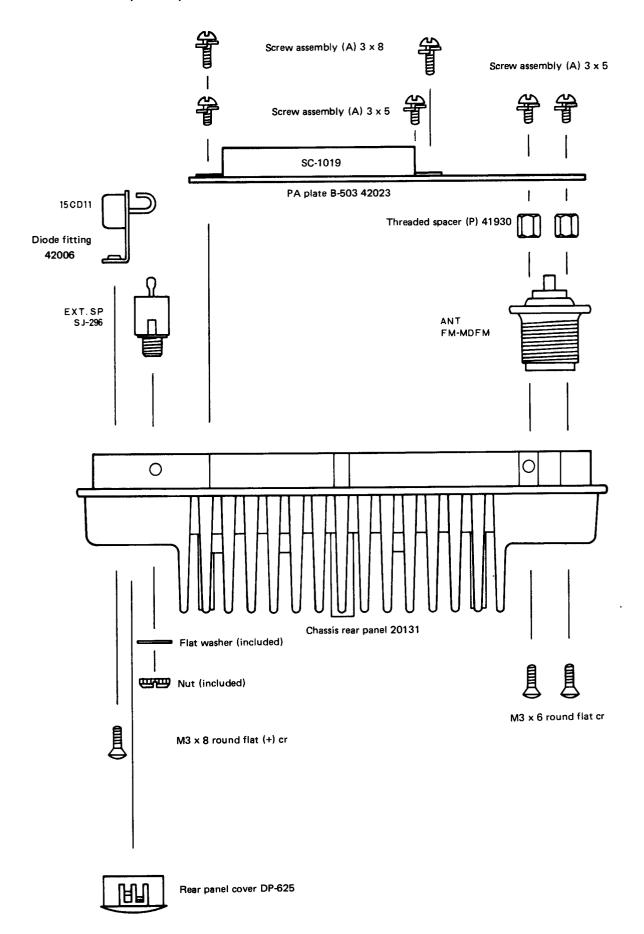


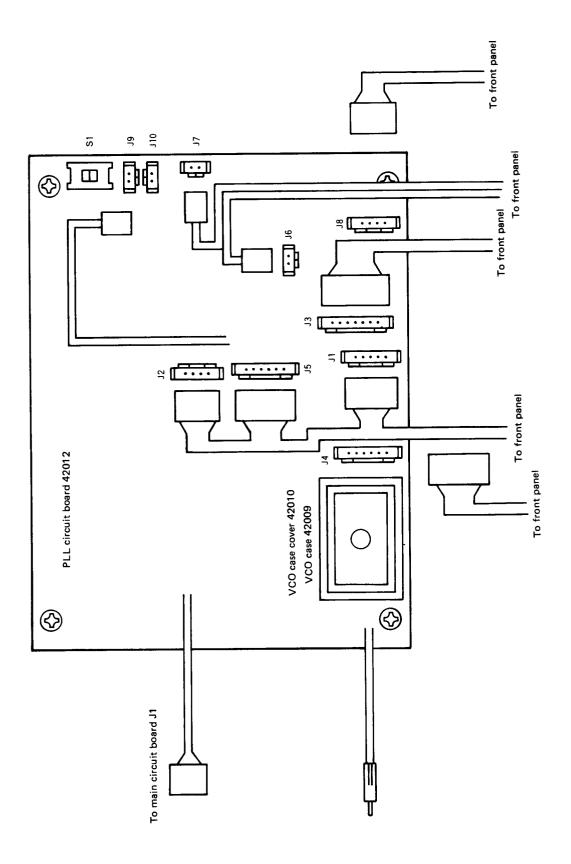
Replacement of front panel controls





Disassembly of rear panel

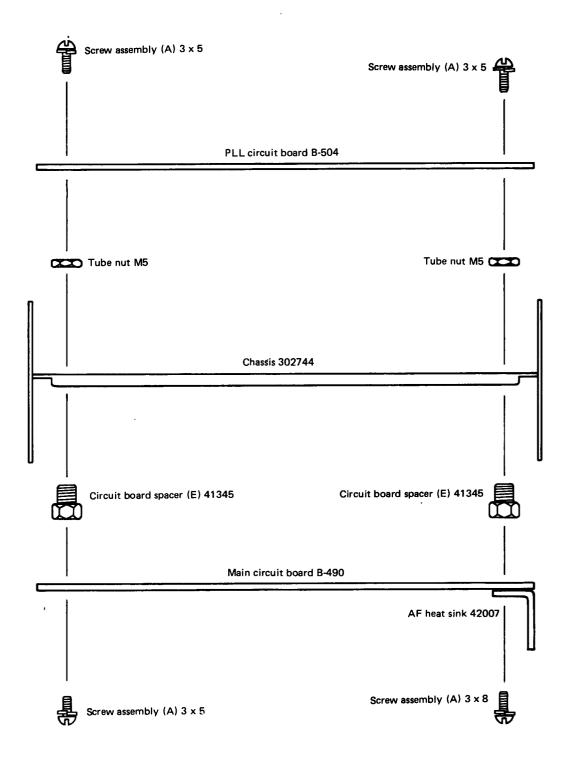




To SP

Main unit wiring

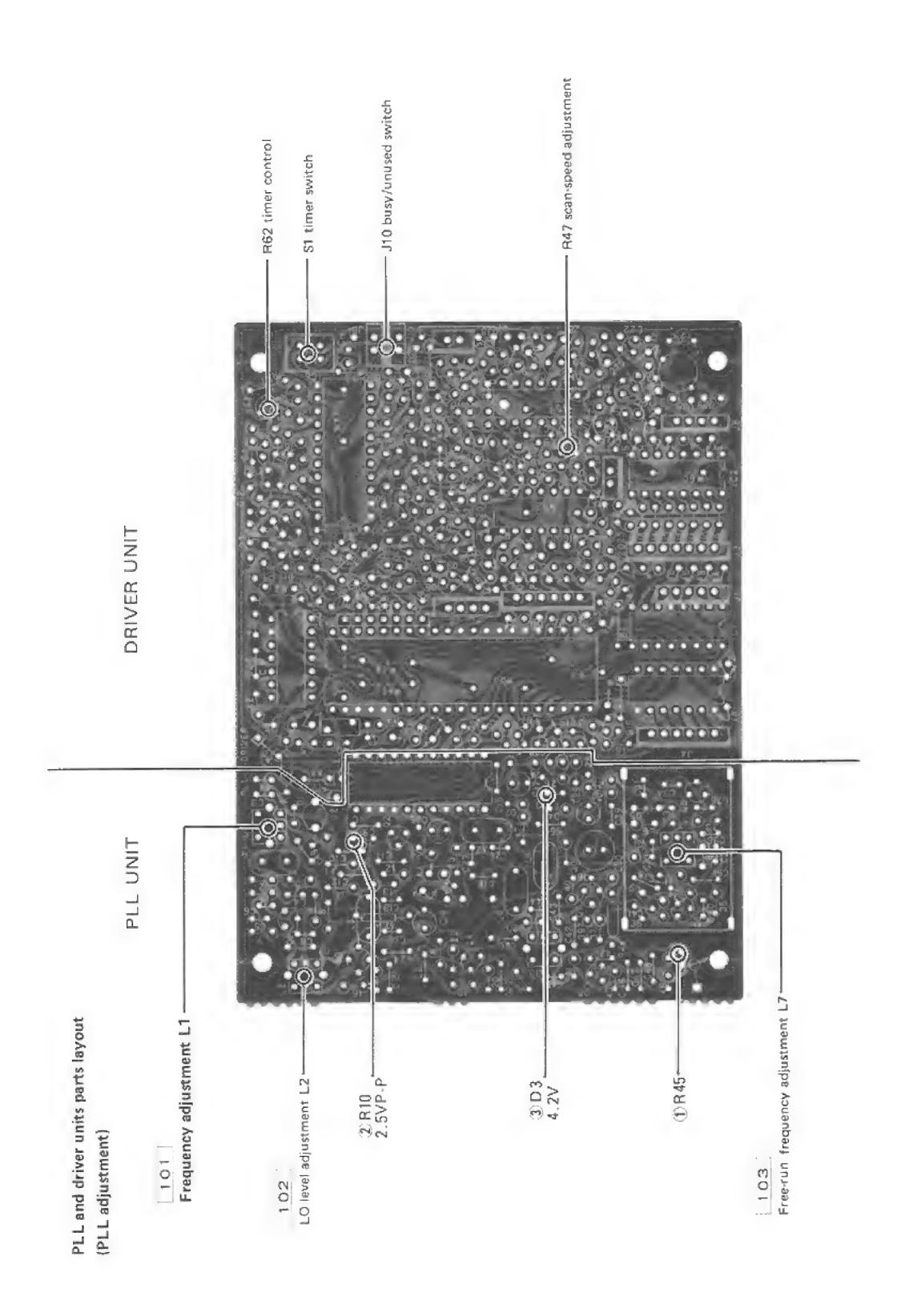
Removal of circuit boards



ADJUSTMENTS

* Adjustment numbers and locations are shown in \bigcirc and \square , and are shown in parts layout drawings.

Notes for PLL adjustment	Adjustment No. Location	
Measuring instruments for adjustment		
Frequency counter (Frequency Range 0.1 \sim 160MHz) Synchroscope (which can measure 10 MHz and above) Multimeter (with internal resistance of approx. 40K Ω /V) RF voltmeter (Frequency Range 0.1 \sim 160MHz) Regulated voltage power supply (DC 13.8V output)		
(Connection of measuring instruments)		
Synchroscope Regulated voltage power supply DC 13.8V		
Frequency counter D3 OR45 Tester		
voltmeter		
1. Frequency adjustment		
 (1) Using the tuning control knob, set the displayed frequency to 145.00MHz. (2) Connect the frequency conter to R45 of the PLL unit. (3) Adjust L1 so that the frequency counter indication is 128.100MHz. 	101	
2. LO level adjustment		
(1) Connect the synchroscope to R1O of the PLL unit.(2) Adjust L2 so that the Amplitude of the waveform becomes 2.7Vp-p or greater.	102	
3. VCO control voltage adjustment		
 Using the tuning control knob, set the displayed frequency to 144.00MHz. Connect the multimeter to the cathode side of D3 of the PLL unit. Adjust L7 so that the voltage at D3 becomes about 4.2V 	103	
4. PLL output level check		
(1) Connect the RF voltmeter to R45 of the PLL unit, and check to be sure that there is 220mV or more.	1	
5. Reference frequency check		
(1) With a reference frequency of 144.00MHz, connect the frequency counter to R45 of the PLL unit, and then check to be sure that the frequency is within the range of $127.1000MHz \pm 200Hz$.		



Measuring instruments for adjustment

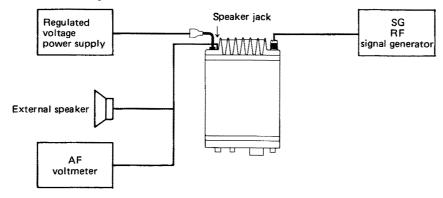
Regulated-voltage power supply (DC 13.8V output)

AF voltmeter (millivoltmeter)

RF signal generator (SG) (Frequency Range 10 ~ 150MHz)

 8Ω external speaker

(Connection of measuring instruments)



1. Reception sensitivity adjustment

- (1) Using the tuning control knob, set to a displayed frequency of 145.00MHz.
- (2) Set an SG output to the extent that the S-meter just begins to move (modulated with 1kHz AF, 7.5kHz deviation)
- (3) Turn L3, L2 and L1 of the main unit, and adjust so that the deflection of the S-meter is maximum.

(Apply the multimeter to the lead wire of R140 of the main unit, and make the adjustment so that the multimeter reading is maximum.)

- (4) Next, set an SG output modulation of 3.5kHz.
- (5) By using L21 of the main unit, make the adjustment so that the S-meter reading is maximum

(Apply the multimeter to the lead wire of R140 of the main unit, and then make the adjustment so that the multimeter reading is maximum.)

Reception sensitivity:

20 dB noise-quieting sensitivity: 0.6µV or more

At $1\mu V$ input S + N/N: 30 dB or more

2. Discriminator adjustment

- (1) Using the tuning control knob, set to a displayed frequency of 145.00MHz.
- (2) Adjust L22 so that AF output becomes maximum.

3. S-meter adjustment

- (1) With SG output at 10 dB, make adjustment of R80 so that four LED's of the S-meter illuminate.
- (2) Next, with the SG output at 40 dB, check to be sure that all of the S-meter LED's are illuminated.

4. Squelch operation check

(1) With the SG output at 10 dB, and with SQL volume at maximum, check to be sure that the squelch opens.

5. AF output check

(1) With the SG output at 10 dB, check to be sure that AF output is 4V or more.

Note: Because there is the possibility, when these adjustments are made, that adjustment of the RF helical cavity may cause band deviation, cross-modulation, etc., it is recommended that the service department of our company be requested to make adjustment of the helical cavity if and when such adjustment becomes necessary.

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7 - 4

Notes for transmitter adjustment

Adjustment
No.
Location

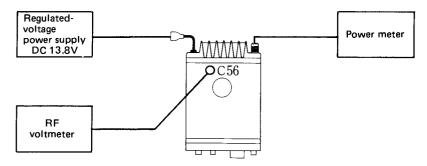
Measuring instruments for adjustment

Power meter (terminated type $130 \sim 160 \text{MHz}$)

RF voltmeter

Regulated-voltage power supply (DC 13.8V output)

(Connections of measuring instruments)



1. Transmission output adjustment

- (1) Using the tuning control knob, set the displayed frequency to 145.00MHz.
- (2) Connect the power meter to the antenna connector.
- (3) Short-circuit R118 and R119 of the main unit, and then disable the ALC. (Also the ALC can be disabled by turning R123 to the right.)
- (4) Adjust L18 and C56 of the main unit so that the power becomes maximum.
- (5) Connect the RF voltmeter to the rotator of C56, and then make adjustment of L15 and L16 so that voltage becomes maximum.
- (6) Once again repeat the adjustment of L18 and C56.

2. Setting the power

- (1) When the power is high, adjust R123 of the main unit to a setting of 25W.
- (2) When the power is low, adjust R116 of the main unit to a setting of approximately 1.2W.

3. RF meter display adjustment

- (1) When the power is low, make adjustment of R110 so that three LED's of the meter (LED) illuminate.
- (2) In this condition, switch to high power, and check to be sure all seven LED's illuminate.

4. APC current check

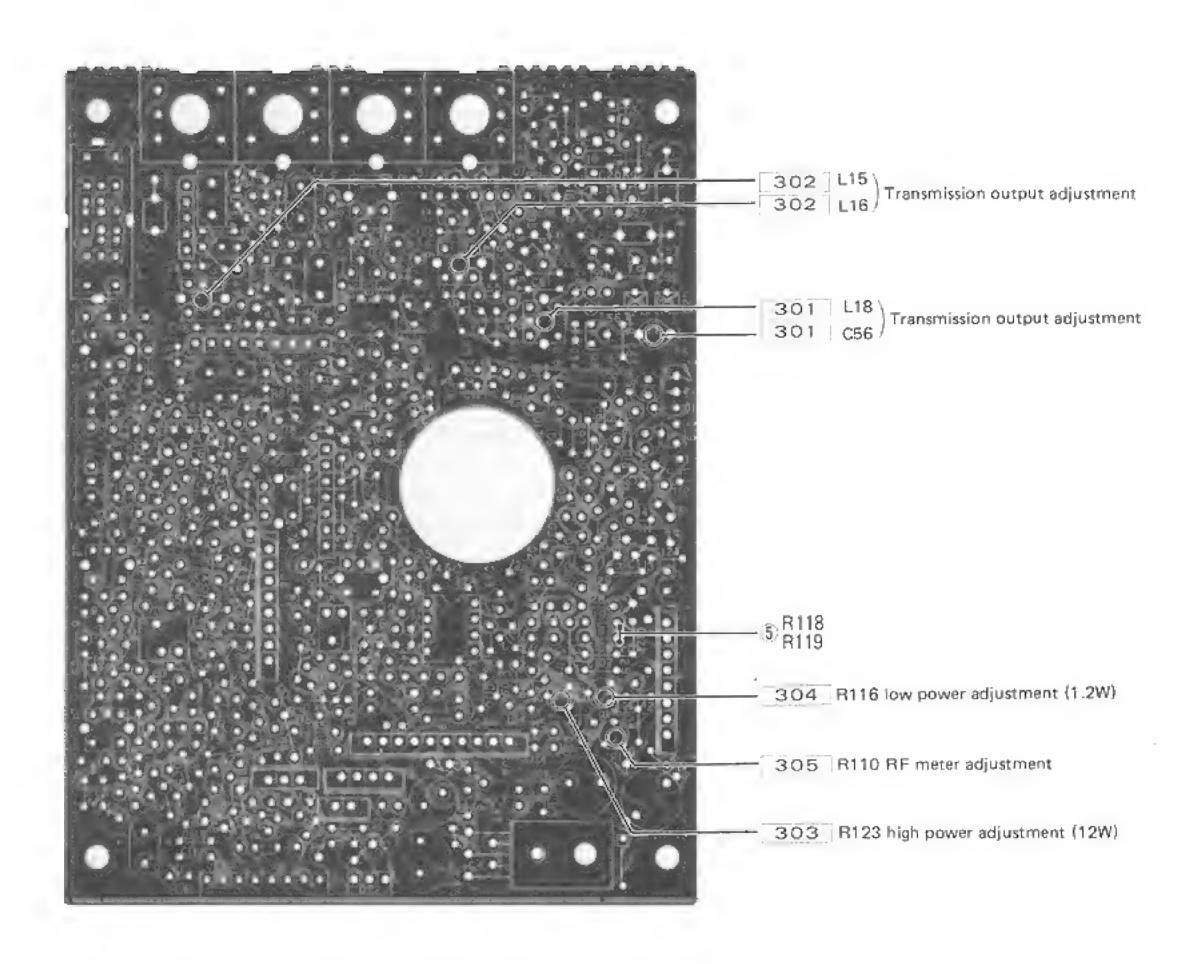
(1) Disconnect the power meter from the antenna connector, measure the current, and check to be sure that the measured reading is equivalent to, or lower than, the transmission current.

(5)

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Adjustment
No.
Location

Measuring instruments for adjustment

Audio generator (AG)

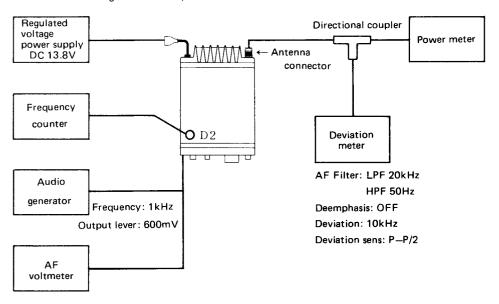
Deviation meter

Frequency counter

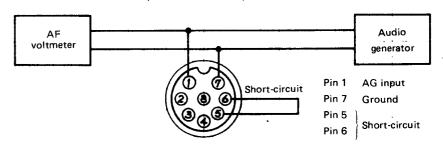
Directional coupler

AF voltmeter (milli-voltmeter)

(Connection of measuring instruments)



Microphone connector (8 pin) connections



1. Local oscillation frequency adjustment

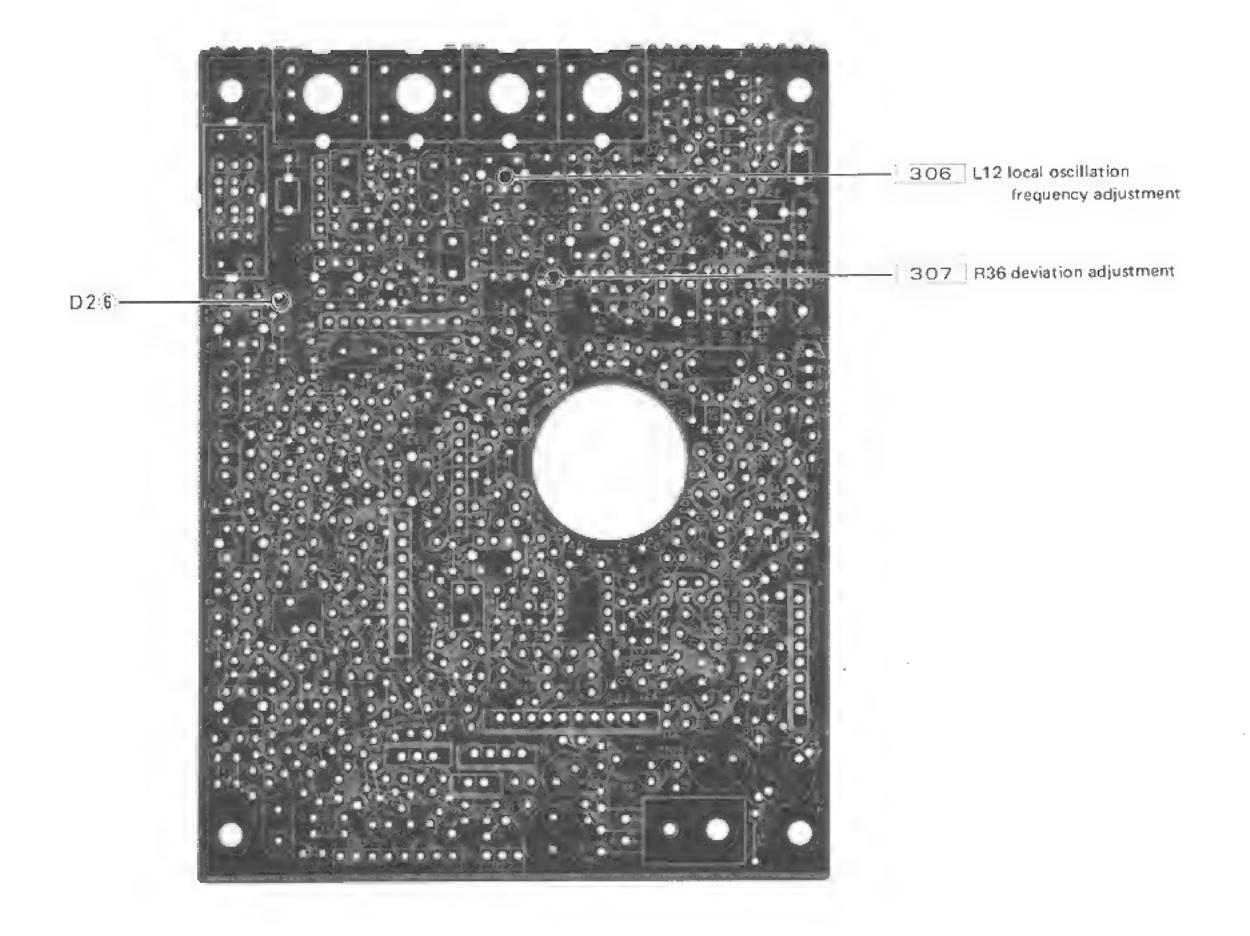
- (1) Connect the frequency counter to D2 of the main unit.
- (2) Make adjustment of L12 of the main unit so that the frequency is within the range of $16.900MHz \pm 100Hz$.

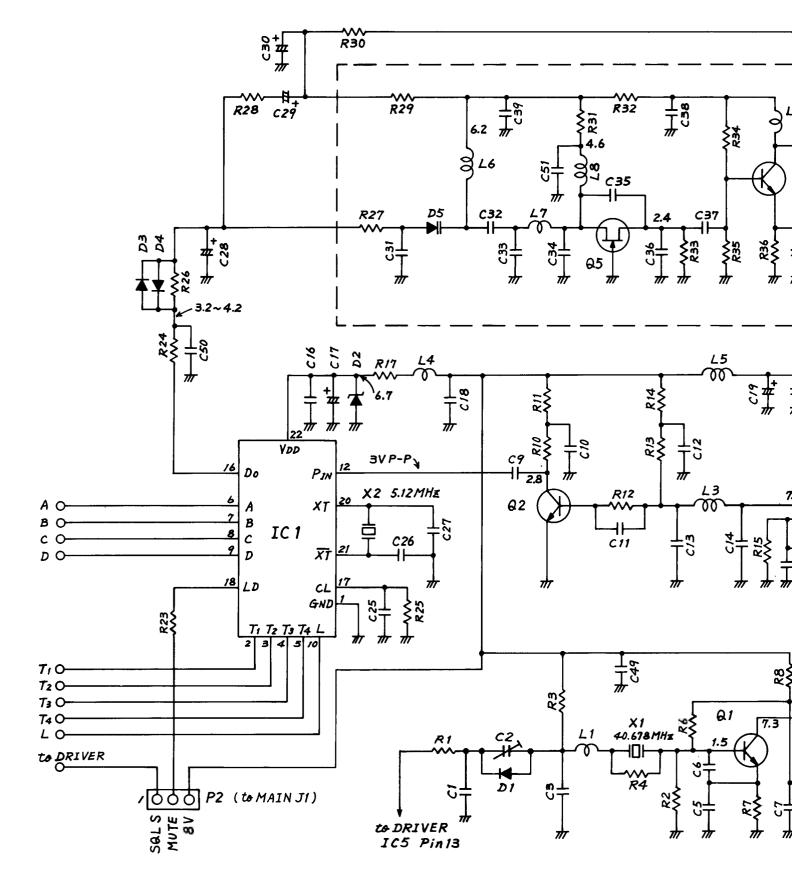
2. Deviation adjustment

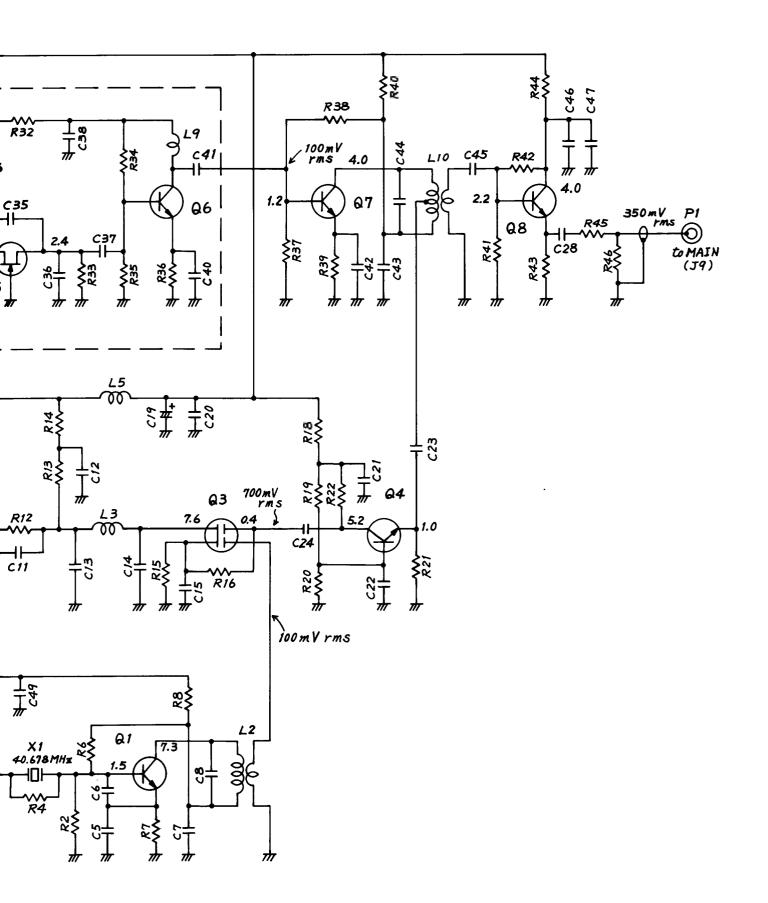
- (1) Input an AG signal (1kHz, 600mV) to the microphone input terminal, and, by using R36 of the main unit, make the adjustment for 4.8kHz ± 0.2kHz.
- . (2) Next, reduce the AG input signal by 20 dB, and, at an input of 60mV, check to be sure that the deviation is then 3.5kHz or more.

6 306

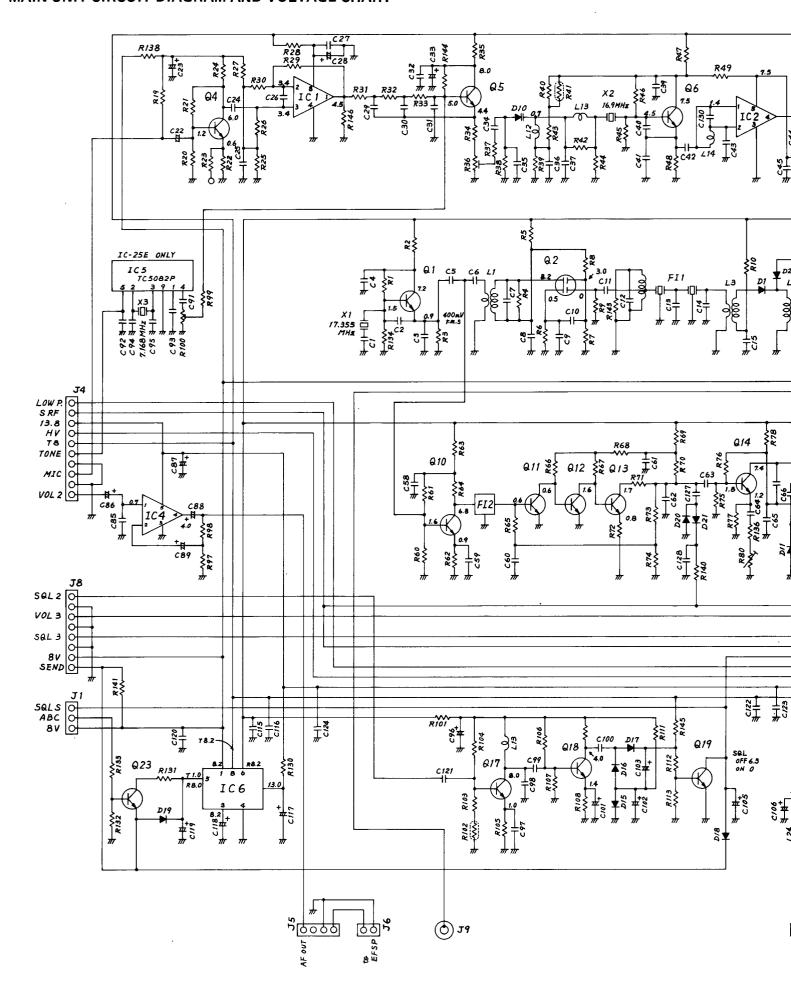
307

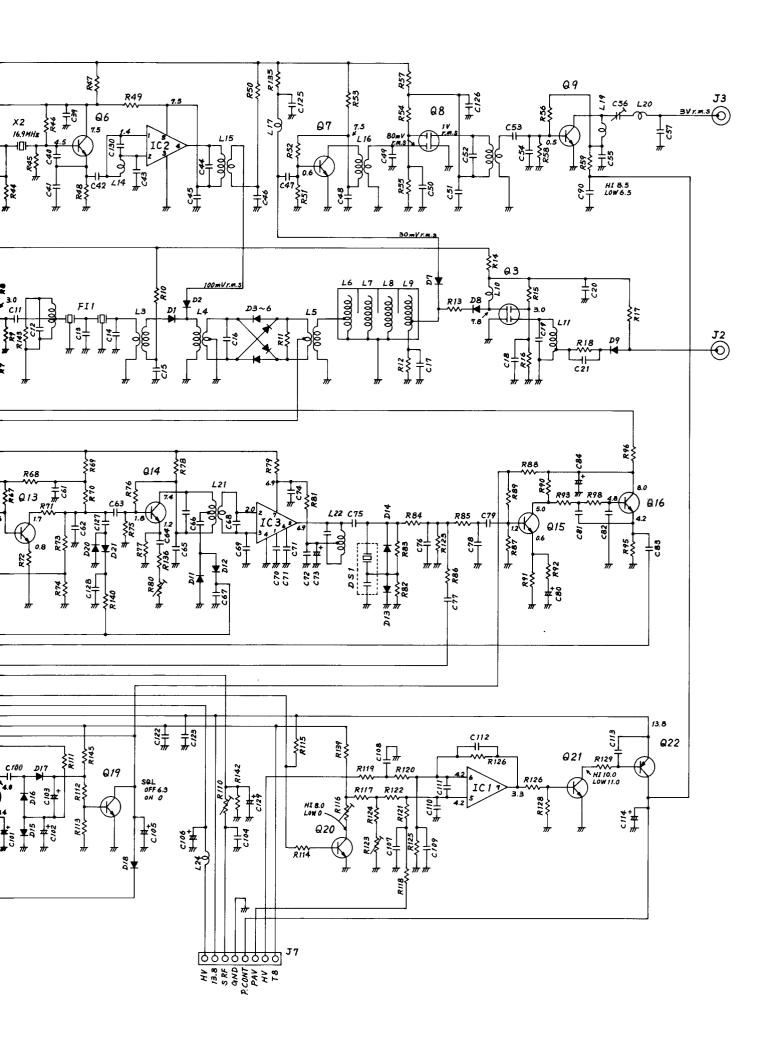


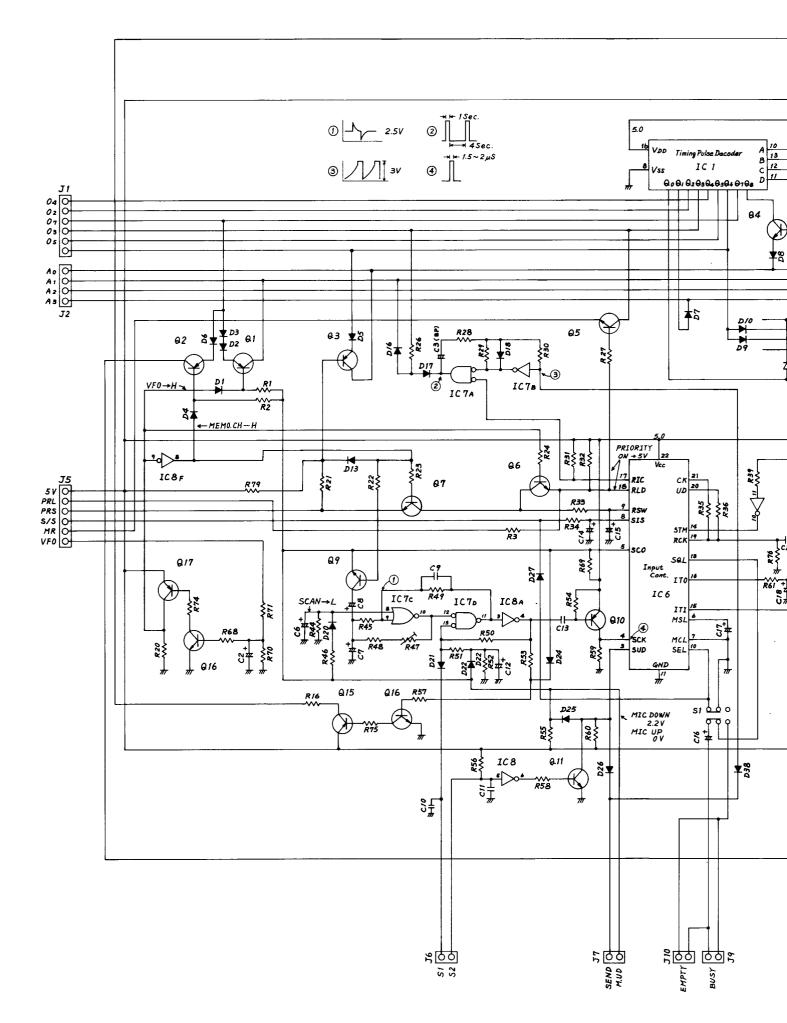


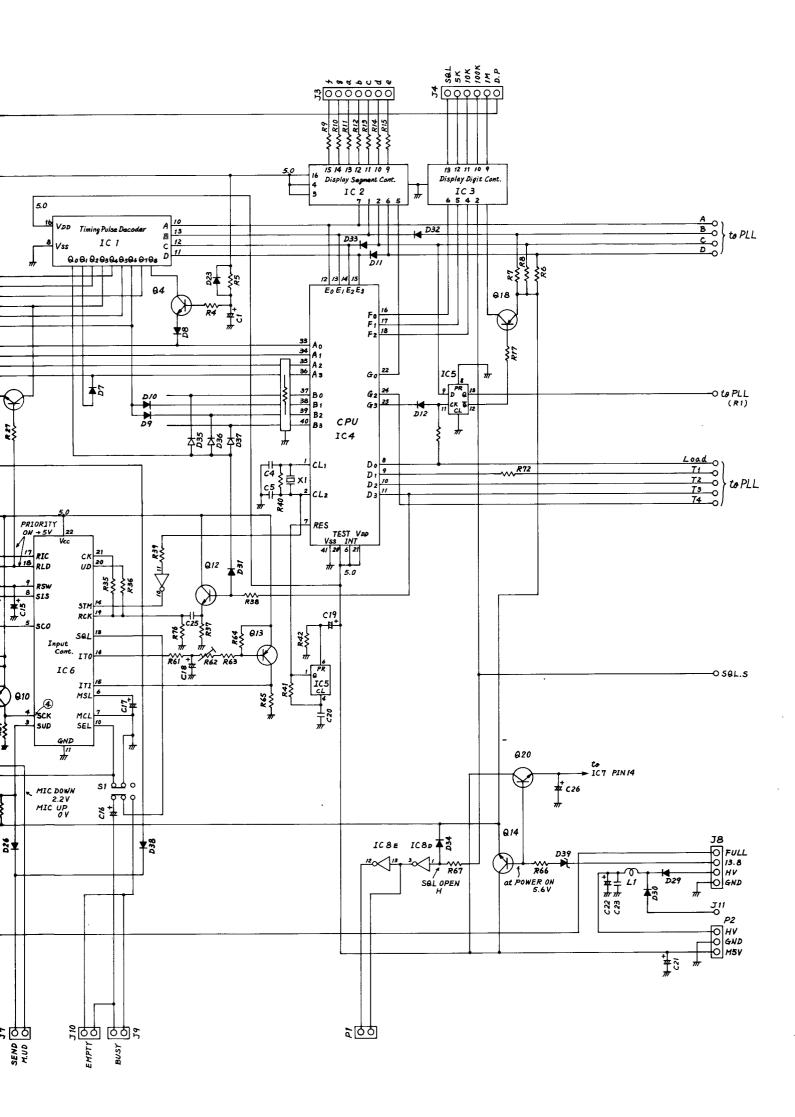


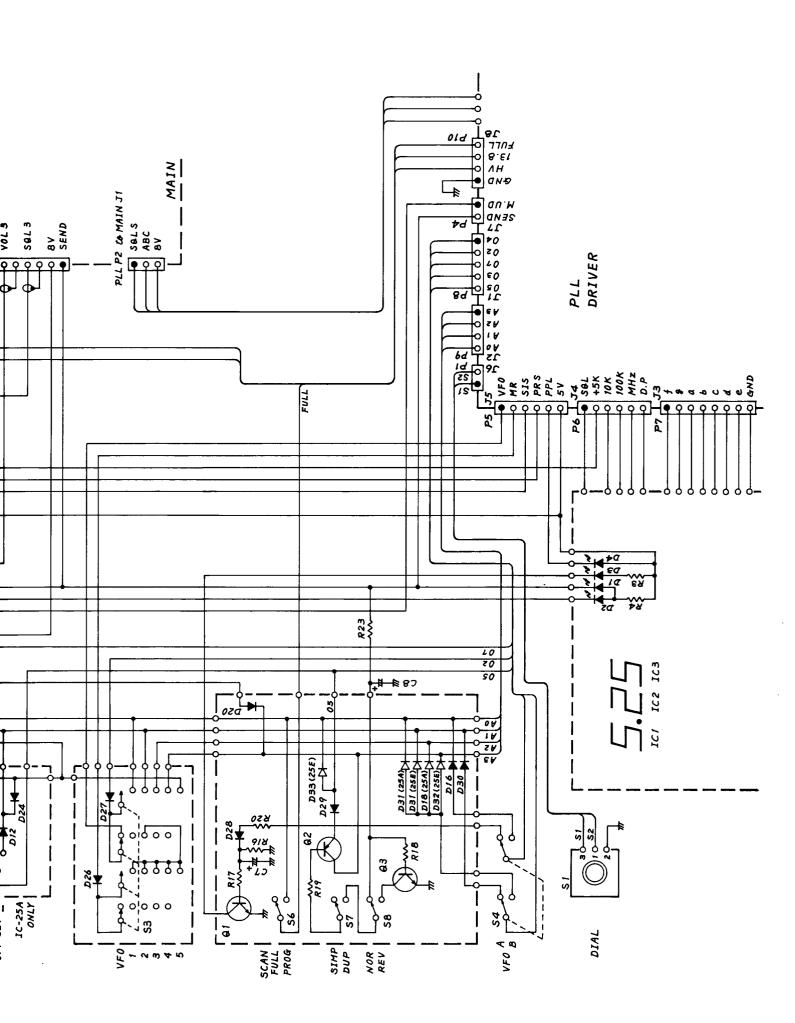
MAIN UNIT CIRCUIT DIAGRAM AND VOLTAGE CHART

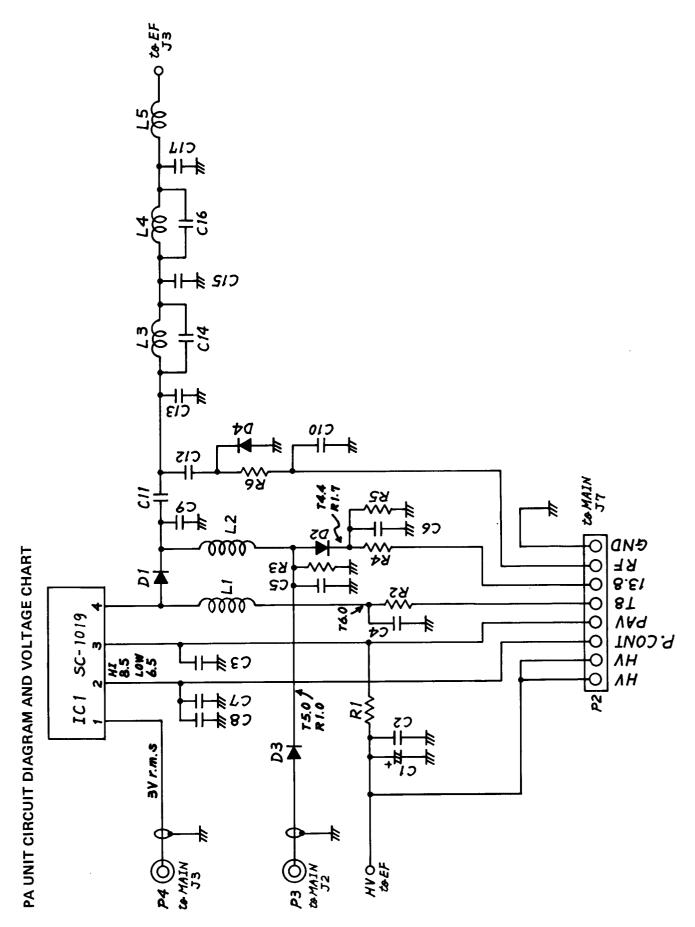




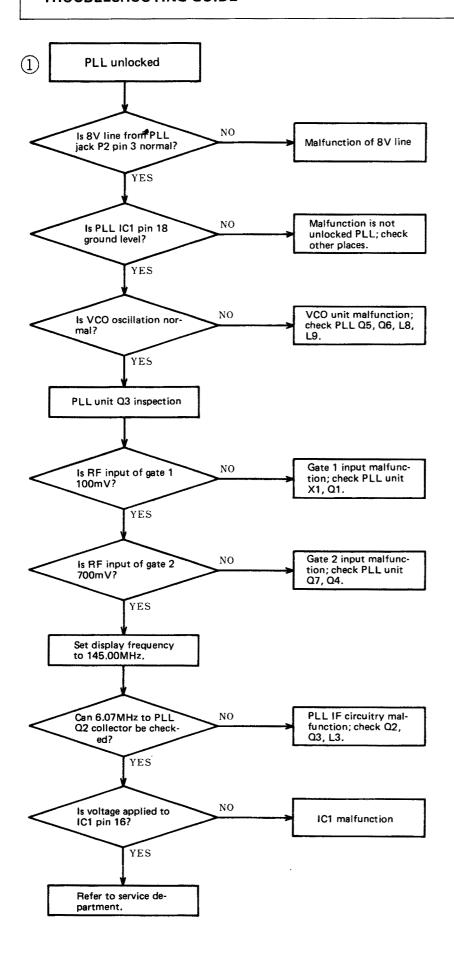


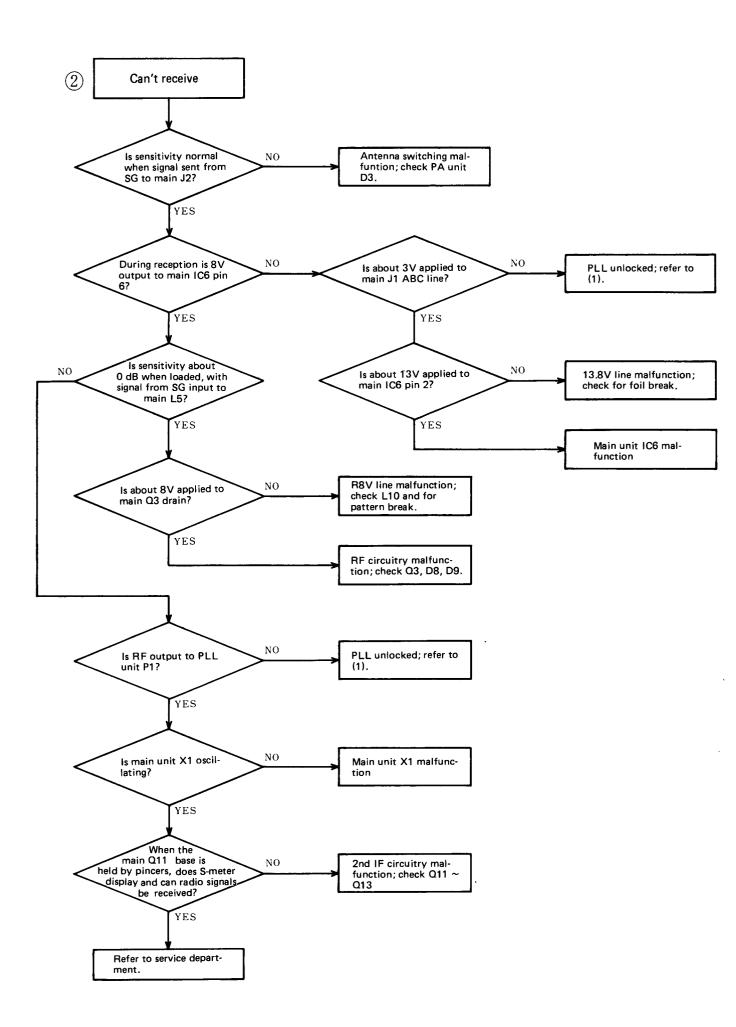


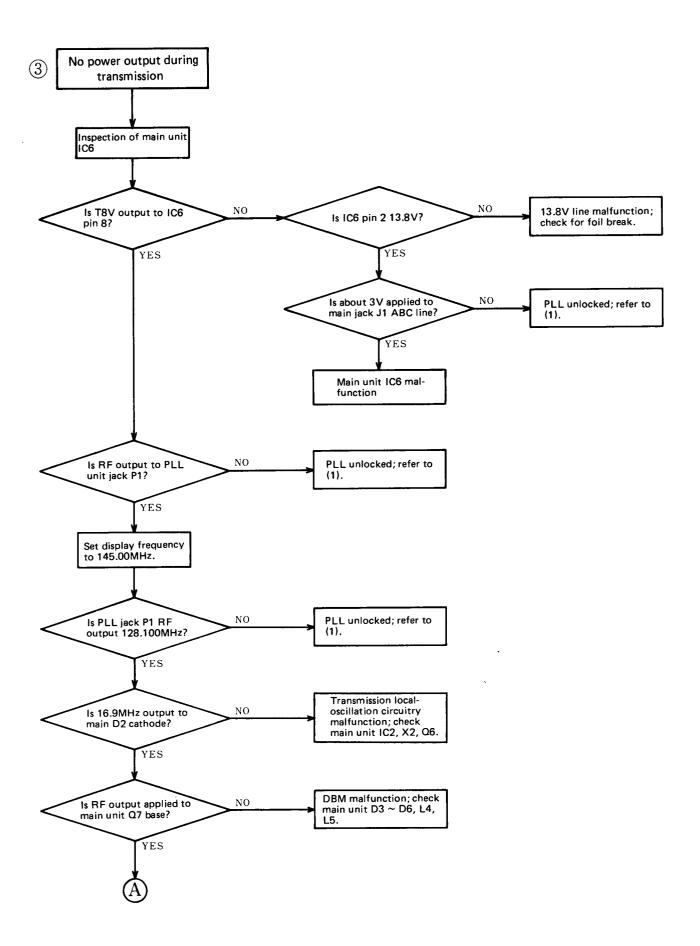


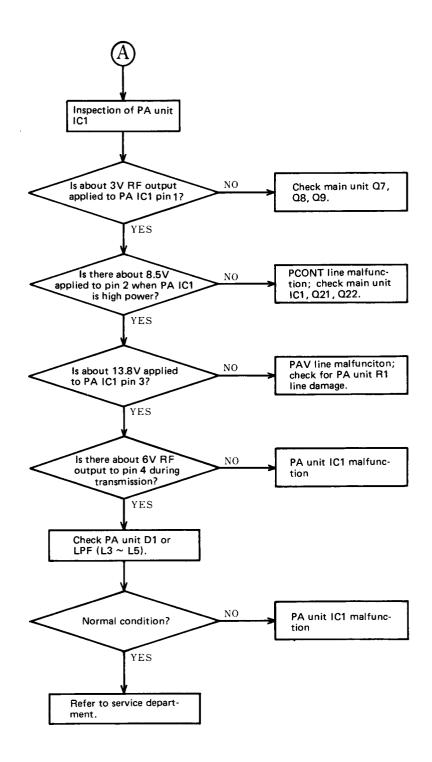


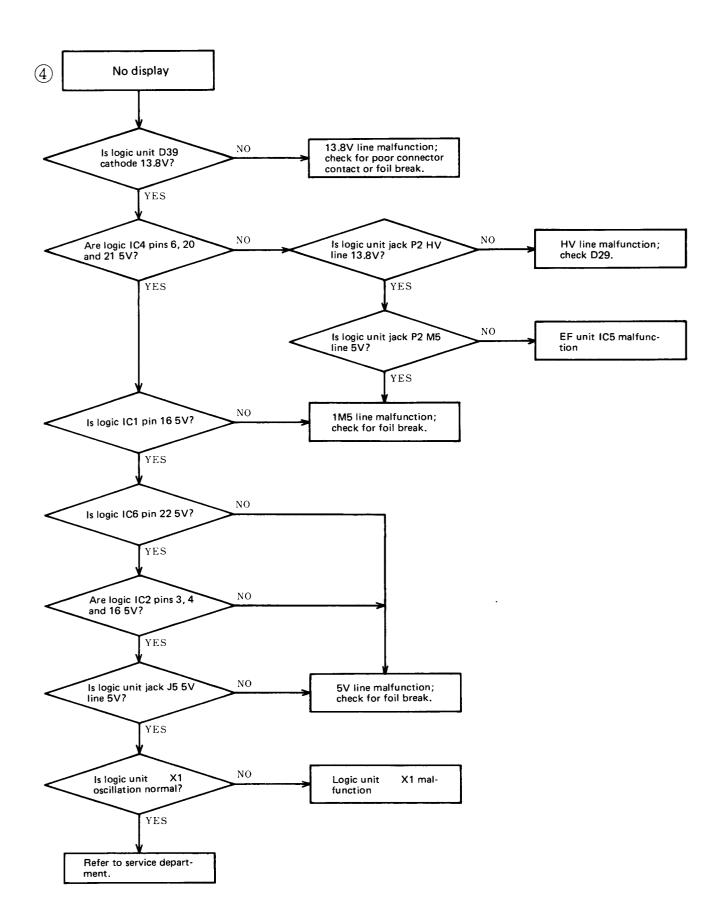
TROUBLESHOOTING GUIDE











PARTS LIST

[EF] UNIT

REF. NO.	DISCRIPTION	PART NO.
IC1 IC2 IC3 IC4 IC5	7SEG. LED 7SEG. LED 7SEG. LED IC IC	TLR312 TLR312 TLR312 TA7612AP 78M05 or 7805
Q1 Q2 Q3	TRANSISTOR TRANSISTOR TRANSISTOR	
D1 D2 D3 D4 D5 D6 D7 D8 D9 D10 D11 D12 D16 D18 D20 D21 D22 D23 D24 D25 D26 D27 D28 D29 D30 D31 D32 D33	LED	SLC-26UR SLC-26GG SLP-451B SLC-26UR LN433YP LN433YP LN433YP LN433YP LN433YP LN233RP LN233RP LN233RP 1SS53 (IC-25A only) 1SS133 1SCD11 GL-9PR4 1SS53 (IC-25A only) 1SS53 1SS53 1SS53 1SS53 1SS53 1SS53 1SS53 1SS133
R1 R2 R3 R4 R5 R6 R7 R8 R9 R10 R11 R12 R13 R14 R15 R16 R17 R18 R19 R20 R21 R22 R23	VARIABLE VARIABLE RESISTOR	K121B1-5N1111 10KA VM13A5M3211 10KB 330 R25 330 R25 1K R25 10 R25 22K R25 10 R25 20K ELR10 47K R10 47K R10 47K R10 47K R10 10K R25 330 R25 22K ELR10

REF. NO.	DISCRIPTION	PART NO.
C1 C2 C3 C4 C5 C6 C7 C8 C9	CERAMIC CERAMIC CERAMIC CERAMIC CERAMIC CERAMIC ELECTROLY TANTALUM CERAMIC CERAMIC	0.001 50V B 0.001 50V B 0.001 50V B 0.001 50V B 0.001 50V B 0.001 50V B 0.47 50V RC2 0.47, 35V, CS15ER47M 0.001 50V B 0.001 50V B
J1 J2 J3	CONNECTOR CONNECTOR CONNECTOR	FM214-8SS(P) (MIC) SJ-296 (EXT SP) FM-MDFM (ANT)
P1 P2 P3 P4 P5 P6 P7 P8 P9 P10 P11 P12 P13 P14 P15	CONNECTOR	TL-25H-10-V1 TL-25H-08-V1 TL-25H-06-V1 TL-25H-06-V1 TL-25H-06-V1 TL-25H-05-V1 TL-25H-04-V1 TL-25H-04-V1 TL-25H-04-V1 TL-25H-04-V1 TL-25H-04-V1 1490-4P TL-25P-02-V1 001T-4100 1P
SP1	SPEAKER	C060A20A000
S1 S2 S3 S4 S5 S6 S7 S8 S9 S10		
B1 B2 B3 B4 B5 B6 B7 B8 B9	P.C BOARD	B-518 (DISP1) B-519 (DISP2) B-520 (SW1) B-521 (SW2) B-522 (SW3) B-525 (SW4) B-526 (SW5) B-523 (METER) B-524 (MIC)

DEE NO	DICCDIDTICN	DADTAO
	DISCRIPTION	
IC1 IC2 IC3 IC4 IC5 IC6 IC7	IC IC IC IC IC IC	4558D BA401 μPC577H μPC2002H TC5082P (IC-25E only) MB3756 ND487C1-3R
Q1 Q2 Q3 Q4 Q5 Q6 Q7 Q8 Q9 Q10 Q11 Q12 Q13 Q14 Q15 Q16 Q17 Q18 Q19 Q20 Q21 Q22 Q23	TRANSISTOR FET FET TRANSISTOR	3SK74M 3SK48 2SC1571G 2SC945 2SC945P 2SC383 3SK74M 2SC2053 2SC763C 2SC945P 2SC945P 2SC945P 2SC945P 2SC945 2SC945 2SC945 ANY RANK 2SC945P 2SC945K 2SC945 ANY RANK 2SC945P 2SC945 ANY RANK 2SC945P 2SC945P 2SC945P
D1 D2 D3 D4 D5 D6 D7 D8 D9 D10 D11 D12 D13 D14 D15 D16 D17 D18 D19 D20 D21 D22	DIODE DIODE DELETED DELETED DELETED DIODE DIODE VARACTOR DIO DIODE	1SS53 1SS53 1SS53 1SS53 1SS53 ODE 1S2688ES 1N60 1N60 1N60 1N60 1N60 1N60 1N60 1N60
FI1 FI2		ER 16M15B1 ER CFU455E2
DS1	DISCRIMINATO	OR CFY-455S
X1 X2 X3	CRYSTAL CRYSTAL CRYSTAL	HC43/U 17.355MHz CR3 HC43/U 7.168MHz (IC-25E only)
L1	COIL	LS216

REF. NO.	DISCRIPTION	PART NO.
REF. NO. L2 L3 L4 L5 L6 L7 L8 L9 L10 L11 L12 L13 L14 L15 L16 L17	COIL COIL COIL COIL COIL COIL COIL COIL	PART NO. LS216 LS216 LR116 LR116 LB83 LB1-1-A LB1-1-A LB14A LW19 LS209 LS-141A LB4 3R6 LB4 100 LS216 LS211 LW-19
L17 L18 L19 L20 L21 L22 L23 L24	COIL COIL COIL COIL COIL CHOKE CHOKE	LS211 LA121 LA121 LS-122 LS-16 102 LW-16
R1 R2 R3 R4 R5 R6 R7 R8 R10 R11 R12 R13 R14 R15 R17 R18 R20 R21 R22 R23 R24 R25 R26 R27 R28 R30 R31 R32 R33 R34 R35	RESISTOR	100K ELR25 1K ELR25 2.2K ELR25 1.2K ELR25 1.2K ELR25 1.00 ELR25 100K ELR25 100K ELR25 100K ELR25 10K ELR25 10K ELR25 10K ELR25 10K ELR25 1K ELR25 1K ELR25 1K ELR25 47 ELR25 47 ELR25 47 ELR25 100K ELR25
R36 R37 R38 R39 R40 R41 R42 R43	TRIMMER RESISTOR RESISTOR RESISTOR RESISTOR THERMISTOR RESISTOR RESISTOR RESISTOR	H0651A 470 15K ELR25 220K ELR25 4.7K ELR25 4.7K ELR25

REF. NO.	DISCRIPTION	PAF	RT NO.
R45	RESISTOR	33K	ELR25
R46	RESISTOR	22K	ELR25
R47	RESISTOR	100	ELR25
R48	RESISTOR	2.7K	ELR25
R49	RESISTOR	47 1K	ELR25 ELR25
R50 R51	RESISTOR RESISTOR	1K	ELR25
R52	RESISTOR	5.6K	ELR25
R53	RESISTOR	100	ELR25
R54	RESISTOR	100K	
R55 R56	RESISTOR RESISTOR	100K 4.7K	ELR25 ELR25
R57	RESISTOR	10	ELR10
R58	RESISTOR	470	ELR25
R59	RESISTOR	22	ELR25
R60	RESISTOR	47K	ELR25
R61 R62	RESISTOR RESISTOR	150K 1K	ELR25 ELR25
R63	RESISTOR	47	ELR25
R64	RESISTOR	1.5K	ELR25
R65	RESISTOR	1.5K	ELR25
R66	RESISTOR	1.5K	ELR25
R67	RESISTOR	1.5K	ELR25
R68 R69	RESISTOR RESISTOR	4.7K 47	ELR25 ELR25
R70	RESISTOR	1K	ELR25
R71	RESISTOR	120	ELR25
R72	RESISTOR	150	ELR25
R73	RESISTOR	10K	ELR25
R74	RESISTOR	3.3K	ELR25
R75 R76	RESISTOR RESISTOR	4.7K 15K	ELR25 ELR25
R77	RESISTOR	10K	ELR25
R78	RESISTOR	2.2K	ELR25
R79	RESISTOR	100	ELR25
R80	TRIMMER	H0651	
R81 R82	RESISTOR RESISTOR	47 10K	ELR25 ELR25
R83	RESISTOR	10K	ELR25
R84	RESISTOR	3.3K	ELR25
R85	RESISTOR	18K	ELR25
R86	RESISTOR	4.7K 39K	R25 ELR25
R87 R88	RESISTOR RESISTOR	4.7K	ELR25
R89	RESISTOR	150K	ELR25
R90	RESISTOR	4.7K	ELR25
R91	RESISTOR	1K	ELR25
R92 R93	RESISTOR RESISTOR	47 5.6K	ELR25 ELR25
R94	RESISTOR	15K	ELR25
R95	RESISTOR	5.6K	ELR25
R96	RESISTOR	100	ELR25
R97	RESISTOR	6.8	ELR25
R98	RESISTOR	100	ELR25
R100	TRIMMER		A 100K E only)
R101	RESISTOR	100	ELR25
R102	THERMISTOR	33D28	1
R103	RESISTOR	2.7K	ELR25
R104	RESISTOR	27K	ELR25
R105 R106	RESISTOR RESISTOR	1K 22K	ELR25 ELR25
R100	RESISTOR	4.7K	ELR25
R108	RESISTOR	1K	ELR25
R109	RESISTOR	4.7K	ELR25
R110	TRIMMER		A 33K
R111 R112	RESISTOR RESISTOR	10K 15K	ELR25 ELR25
2	1.25/07/5/1		

REF. NO.	DISCRIPTION	PART NO.
R113	RESISTOR	22K ELR25
R114	RESISTOR	100K ELR25
R115	RESISTOR	100K ELR25
R116	TRIMMER	H0651A 3.3K
R117	RESISTOR	150K ELR25
R118	RESISTOR	1K R25 1K R25
R119	RESISTOR RESISTOR	47K ELR25
R120 R121	RESISTOR	47K ELR25
R122	RESISTOR	15K ELR25
R123	TRIMMER	H0651A 3.3K
R124	RESISTOR	5.6K ELR25
R125	RESISTOR	22K ELR25
R126	RESISTOR RESISTOR	1M ELR25 10K ELR25
R127 R128	RESISTOR	2.2K ELR25
R129	RESISTOR	330 ELR25
R130	JUMPER	JPW-02H
R131	RESISTOR	15K ELR10
R132	RESISTOR	10K ELR10
R133	RESISTOR	22K ELR10
R134	RESISTOR	47K ELR25
R135	RESISTOR RESISTOR	100 ELR25 100 ELR25
R136 R137	RESISTOR	47K ELR25
R138	RESISTOR	100 ELR25
R139	RESISTOR	1.8K ELR25
R140	RESISTOR	22K ELR25
R141	RESISTOR	2.2K R25
R142	RESISTOR	3.9K R25
R143	RESISTOR	6.8K ELR25
R144	RESISTOR	47K ELR25 47K ELR25
R145 R146	RESISTOR RESISTOR	10K ELR25
R146	RESISTOR	1K ELR10
1147	1120101011	(IC-25E only)
R148	RESISTOR	10K ELR25
		(IC-25E only)
R149	JUMPER	JPW-02A
R151 R152	JUMPER JUMPER	JPW-02H JPW-02H
N 102	JUNIFER	JF W-02F1
C1	DIP MICA	30P 50V
C2	DIP MICA	100P 50V
C3	DIP MICA	100P 50V
C4	CERAMIC	0.0047 50V B
C5 C6	CERAMIC CERAMIC	5P 50V SL 0.001 50V B
C7	CERAMIC	10P 50V SL
C8	CERAMIC	0.0047 50V B
C9	CERAMIC	0.0047 50V B
C10	CERAMIC	0.0047 50V B
C11	CERAMIC	0.001 50V B
C12	CERAMIC	10P 50V SL
C13 C14	CERAMIC CERAMIC	20P 50V SL 10P 50V SL
C14	CERAMIC	0.0047 50V B
C16	CERAMIC	10P 50V SL
C17	CERAMIC	220P 50V SL
C18	CERAMIC	0.0047 50V B
C19	CERAMIC	2P 50V SL
C20	CERAMIC	0.0047 50V B 33P 50V SL
C21 C22	CERAMIC ELECTROLY	33P 50V SL 1 50V B.P
C22	ELECTROLY	100 10V RE or MS
C24	CERAMIC	0.001 50V B
C25	BARRIER LAY	
C26	CERAMIC	0.001 50V B

REF. NO.	DISCRIPTION	PAR	T NO.
C27	CERAMIC	0.0047	50V B
C28	ELECTROLY		10V
C29	BARRIER LAY		TBD05X103
C30	BARRIER LAY		TBD05X103
C31	MYLAR	0.0022	
C32	CERAMIC	0.0047	
C33	ELECTROLY		10V
C34	BARRIER LAY		
C35	MYLAR	0.0047	
C36	BARRIER LAY		
C39		0.0047	
C40	CERAMIC CERAMIC CERAMIC	100P	
C41	CERAMIC	100P	50V YL
C42	CERAMIC	10P	50V SL
C43	CERAMIC	0.0047	
C44	CERAMIC	10P	
C45	CERAMIC	0.0047	
C46	CERAMIC	0.0047	
C47	CERAMIC	2P	50V SL
C48	CERAMIC	0.0047	50V B
C49	CERAMIC	2P	50V SL
C50	CERAMIC	0.0047	50V B
C51	CERAMIC	220P	
C52	CERAMIC	3P	50V SL
C53	CERAMIC	7P	50V SL
C54	CERAMIC	47P	
C55	CERAMIC		
C56	TRIMMER		
C57	CERAMIC		
C58	BARRIER LAY	0.1	25V
C59	BARRIER LAY		25V
C60	BARRIER LAY		25 V
C61	BARRIER LAY		25V
C62	MYLAR	0.001	
C62	MYLAR	0.001	
C63	BARRIER LAY		
C65	BARRIER LAY		25V
	CERAMIC	220P	50V SL
C66	CERAMIC	0.001	50V B
C67 C68	MYLAR	0.0022	
C69	BARRIER LAY		25V
C70	BARRIER LAY		25V
C70	BARRIER LAY		25V
C71	BARRIER LAY		25V
C72	ELECTROLY	4.7	10V RC2
C73	TANTALUM	2.2	16V 1162
C74 C75	MYLAR	0.001	50V
C75	MYLAR	0.001	50V
C77	BARRIER LAY		TBD05X103
C77	BARRIER LAY		35V
C79	MYLAR	0.0047	
C80	ELECTROLY	47	10V RE or MS
C81	BARRIER LAY		TBD05X103
C82	BARRIER LAY		TBD05X103
C83	BARRIER LAY		25V
C84	ELECTROLY	10	10V RC2
C85	CERAMIC	470P	50V B
C86	ELECTROLY	1	10V RE or MS
C87	ELECTROLY	-	16V RE or MS
C88	ELECTROLY	220	10V RE or MS
C89	ELECTORLY	100	10V RE or MS
C90	CERAMIC	0.0047	
C90 C91	BARRIER LAY		25V
Cal		(IC-25E	
C92	BARRIER LAY	-	-
USZ		(IC-25E	
C93	CERAMIC	470P	50V B
1,4.4		-	

REF. NO.	DISCRIPTION	PAR	T NO.
C94	CERAMIC	47P	50V SL
C95	CERAMIC	(IC-25E 47P	50V SL
C96	ELECTROLY	(IC-25E 2.2	only) 50V
C97	BARRIER LAY		25V
C98	BARRIER LAY		35V
C99 C100	MYLAR BARRIER LAY	0.0022 0.047	50V 35V
C100	ELECTROLY		10V RC2
C102	ELECTROLY	10	10V RC2
C103	ELECTROLY CERAMIC	3.3 0.0047	10V RC2 50V B
C104 C105	ELECTROLY	2.2	10V RC2
C106	ELECTROLY	470	16V MS or RE
C107	CERAMIC	220P	50V SL
C108 C109	CERAMIC CERAMIC	220P 0.0047	50V 50V B
C110	CERAMIC	0.0047	1
C111	CERAMIC	0.001	50V B
C112	CERAMIC CERAMIC	0.001 0.0047	50V B
C113 C114	ELECTROLY		10V RC2
C115	BARRIER LAY	0.1	25V
C116	BARRIER LAY BARRIER LAY		25V 16V
C118 C119	BARRIER LAY	1	50V
C120	BARRIER LAY	0.1	25V
C121	BARRIER LAY		TBD05X103
C122	CERAMIC CERAMIC	220P 470P	50V SL 50V B
C123 C124	CERAMIC	220P	50V SL
C125	CERAMIC	220P	50V SL
C126	CERAMIC	220P	50V SL
C127 C128	CERAMIC CERAMIC	470P 0.001	50V B 50V B
C128	ELECTROLY	10	10V RC2
C130	CERAMIC	15P	50V SL
C131 C132	CERAMIC BARRIER LAY	220P 0.1	50V SL 25V
J1 J2	CONNECTOR	TL-25P	
J3	CONNECTOR	TMP-J0	
J4	CONNECTOR	TL-25P	
J5	CONNECTOR CONNECTOR	TL-25P	
J6 J7	CONNECTOR	TL-25P	
J8	CONNECTOR	TL-25P	-08-V1
J9	CONNECTOR	TMP-J0	1X-A
B1	P.C BOARD	B-490	
			:
			:

[PLL] UNIT

REF. NO.	DISCRIPTION	PART NO.
IC1	IC	TC9123 BP
Q1	TRANSISTOR	
Q2 Q3	TRANSISTOR FET	2SC763C 3SK74M
Q4	TRANSISTOR	
Q5	FET	2SK125
Q6	TRANSISTOR	
Q7 Q8	TRANSISTOR TRANSISTOR	
Co .	INANSISTON	230303 1101
D1	DIODE	1SS53
D2 D3	ZENER DIODE	XZ068 1S953
D3	DIODE	1S953
D5	VARACTOR DI	
L1	COIL	LS-134
L2	COIL	LS-145
L3	CHOKE	5R6
L4	CHOKE	101
L5 L6	CHOKE COIL	101 LW-19
L7	COIL	LB-73
L8	COIL	LW-19
L9	COIL	LW-19
L10	COIL	LS-211
X1	CRYSTAL	HC43/U 40.678MHz
X2	CRYSTAL	HC43/U 5.12MHz
R1	RESISTOR	4.7K R25
R2 R3	RESISTOR RESISTOR	2.7K ELR25 4.7K ELR25
R4	RESISTOR	1.0K ELR25
R5	RESISTOR	4.7K ELR25
R6	RESISTOR	18K ELR25
R7	RESISTOR	1.2K ELR25
R8 R10	RESISTOR RESISTOR	1K ELR25 470 R25
R11	RESISTOR	100 ELR25
R12	RESISTOR	68K ELR25
R13	RESISTOR	470 ELR25
R14 R15	RESISTOR RESISTOR	100 ELR25 470 ELR25
R16	RESISTOR	1K ELR25
R17	RESISTOR	47 ELR25
R18	RESISTOR	100 ELR25
R19 R20	RESISTOR RESISTOR	22K ELR25 5.6K ELR25
R21	RESISTOR	220 ELR25
R22	RESISTOR	1K ELR25
R23	RESISTOR	47K ELR25
R24 R25	RESISTOR RESISTOR	2.2K ELR25 1K ELR25
R26	RESISTOR	15K ELR25
R27	RESISTOR	470 ELR25
R28	RESISTOR	470 ELR25
R29 R30	RESISTOR RESISTOR	2.2 ELR25 150 ELR25
R30 R31	RESISTOR	220 ELR25
R32	RESISTOR	220 ELR25
R33	RESISTOR	330 ELR25
R34 R35	RESISTOR RESISTOR	5.6K ELR25 1.2K ELR25
R36	RESISTOR	33 ELR25

REF. NO.	DISCRIPTION	PAF	RT NO.
R37	RESISTOR	1.2K	ELR25
R38	RESISTOR	5.6K	ELR25
R39		- 33	ELR25
R40	RESISTOR	47	ELR25
R41	RESISTOR	2.7K	ELR25
R42	RESISTOR	1.8K	ELR25
R43	RESISTOR	100	ELR25
R44	RESISTOR RESISTOR	220	R25
R45 R46	RESISTOR	22 120	R25 ELR25
R47	JUMPER	JPW-02	
C1	CERAMIC	68P	
C2	TRIMMER	CV05D	
C3	CERAMIC	10P	50V SL
C4	CERAMIC	0.001	
C5	CERAMIC	47P	50V SL
C6	CERAMIC	33P	50V SL
C7 C8	CERAMIC CERAMIC	0.0047 7P	50V B 50V SL
C9	BARRIER LAY		
C10	CERAMIC	0.0047	
C11	CERAMIC		50V B
C12	CERAMIC		50V B
C13	CERAMIC	33P 33P	50V SL
C14 C15	CERAMIC CERAMIC	0.0047	
C15	BARRIER LAY		25V
C17	ELECTROLY	47	10V
C18	CERAMIC	0.0047	50V B
C19	ELECTROLY	100	10V
C20	CERAMIC	220P	
C21	CERAMIC	0.0047	
C22 C23	CERAMIC CERAMIC	0.0047 10P	50V B 50V SL
C23	CERAMIC	0.0022	
C25	ELECTROLY	2.2	50V
C26	DIP MICA	39P	50V
C27	DIP MICA	39P	50V
C28	TANTALUM	0.47 10	35V
C29 C30	ELECTROLY ELECTROLY	220	16V 10V
C31	CÉRAMIC	470P	50V B
C32	CERAMIC	50P	50V XL
C33	CERAMIC	5P	50V CH
C34	CERAMIC	3P	50V SL
C35 C36	CERAMIC CERAMIC	15P 8P	50V SL 50V SL
C37	CERAMIC	1P	50V SL
C38	CERAMIC	220P	50V SL
C39	CERAMIC	0.001	50V B
C40	CERAMIC	470P	50V B
C41	CERAMIC CERAMIC	47P 0.0047	50V SL 50V B
C42 C43	CERAMIC	0.0047 220P	50V B 50V SL
C43	CERAMIC	3P	50V SL
C45	CERAMIC	22P	50V SL
C46	CERAMIC	0.0047	50V B
C47	CERAMIC	220P	50V SL
C48	CERAMIC CERAMIC	0.0047 0.0047	50V B 50V B
C49 C50	CERAMIC	0.0047	
C50 C51	CERAMIC	0.0047	
C52	BARRIER LAY		25V
P1	CONNECTOR	TMP-P0	1X-A1
P2	CONNECTOR	TL-25H	
В1	P.C BOARD	B-504	
1			

[DRIVER] UNIT

REF.NO.	DISCRIPTION	PART NO.
IC1 IC2 IC3 IC4 IC5 IC6 IC7	IC IC IC IC IC IC IC	4028 4511 M54516 μPD650-108 4013 MB14025 4001 (UBP) (C) 4069
Q1 Q2 Q3 Q4 Q5 Q6 Q7 Q9 Q10 Q11 Q12 Q13 Q14 Q15 Q16 Q17 Q18 Q19 Q20	TRANSISTOR TRANSISTOR	2SA1015 2SC945 ANY RANK 2SC945 ANY RANK 2SA30 or 831 2SD468 2SA1015 2SC945 ANY RANK 2SA1015 2SC945 ANY RANK 2SA1015 2SC945 ANY RANK
D1 D2 D3 D4 D5 D6 D7 D8 D9 D10 D11 D12 D13 D16 D17 D18 D19 D20 D21 D22 D23 D24 D25 D26 D27 D29 D30 D31 D32 D33 D34 D35 D36 D37 D38 D39 D41	DIODE	1S953 1SS53 1SS53 1S953 1SS53

REF. NO.	DISCRIPTION	PART NO.
D42 D43	DIODE DIODE	1SS53 (1SS133) 1SS53 (1SS133)
L1	СНОКЕ	LW-12
X1	CERAROCK	CSB400A
S1	SWITCH	SSS022
R53 R54 R55	RESISTOR RESISTOR RESISTOR	100K ELR25 1K ELR25 680 ELR25
R56 R57	RESISTOR RESISTOR RESISTOR	100K ELR25 47K ELR25
R58 R59	RESISTOR RESISTOR	22K ELR25 1K ELR25
R60 R61	RESISTOR RESISTOR	22K ELR25 1K ELR10

[PA] UNIT

REF. NO.	DISCRIPTION	PART NO.
R62	TRIMMER	H0651A 220K
R63	RESISTOR	10K ELR25
R64	RESISTOR	1M ELR25
R65	RESISTOR	1.8K ELR25
R66	RESISTOR	1K ELR25
R67	RESISTOR	47K ELR25
R68	RESISTOR RESISTOR	47K ELR25 4.7K ELR25
R69 R70	RESISTOR	100K ELR25
R71	RESISTOR	10K ELR25
R72	RESISTOR	1K ELR25
R73	RESISTOR	150 ELR25
R74	RESISTOR	22K ELR25
R75	RESISTOR	22K ELR25
R76	RESISTOR	680 ELR25
R79	RESISTOR	47K ELR25
R80 R81	JUMPER JUMPER	JPW-02H JPW-02H
R82	JUMPER	JPW-02A
R83	JUMPER	JPW-02A
R84	JUMPER	JPW-02H
R85	JUMPER	JPW-02H
R86	JUMPER	JPW-02H
R87	JUMPER	JPW-02H
R88	JUMPER	JPW-02H
R89	RESISTOR	390 ELR25
C1	ELECTROLY	10 10V RC2
C2	ELECTROLY	0.47 50V RC2
C3	ELECTROLY	4.7 BP 25V
C4 C5	CERAMIC	100P 50V SL
C6	CERAMIC ELECTROLY	100P 50V SL 4.7 50V RC2
C7	ELECTROLY	4.7 50V RC2
C8	ELECTROLY	4.7 25VBP
C9	CERAMIC	0.0022 50V B
C10	BARRIER LAY	· · - · · · - · · · · ·
C11	CERAMIC	0.001 50V B
C12	ELECTROLY	0.1 50V RC2
C13 C14	CERAMIC	0.001 50V B 0.1 50V RC2
C14	ELECTROLY ELECTROLY	0.1 50V RC2 0.1 50V RC2
C16	DELETED	0.1 307 1102
C17	ELECTROLY	0.47 50V RC2
C18	ELECTROLY	47 10V RE or CE
C19	ELECTROLY	0.47 50V RC2
C20	BARRIER LAY	
C21	ELECTROLY	100 10V
C22	ELECTROLY	470 16V
C23 C25	BARRIER LAY BARRIER LAY	
C25	TANTALUM	47 10V
14	CONNECTOR	TI 250 05 1/1
J1 J2	CONNECTOR CONNECTOR	TL-25P-05-V1 TL-25P-04-V1
J3	CONNECTOR	TL-25P-04-V1
J4	CONNECTOR	TL-25P-06-V1
J5	CONNECTOR	TL-25P-06-V1
J6	CONNECTOR	TL-25P-02-V1
J7	CONNECTOR	TL-25P-02-V1
J8 J9	CONNECTOR CONNCETOR	TL-25P-04-V1 TL-25P-02-V1
J9 J10	CONNCETOR	TL-25P-02-V1
J11	CONNECTOR	RT-01T-10B (1P)
J12	CONNECTOR	RT-01T-1.0B
P1	CONNECTOR	SMP03VB
P2	CONNECTOR	TL-25H-02-A1
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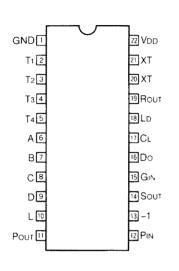
REF. NO.	DISCRIPTION	PART NO.	
IC1	IC	SC1019	
D1 D2 D3 D4	DIODE DIODE DIODE DIODE	M1402 M1402 M1301 1SS97	
L1 L2 L3 L4 L5 L6	CHOKE COIL COIL COIL COIL	LW-19 LA-127 LA-2 LA-2 LA-71 LW-19	
R1 R2 R3 R4 R5	RESISTOR RESISTOR RESISTOR RESISTOR RESISTOR RESISTOR	0.15 2W 120 ELR25 1K ELR25 1.5K ELR25 220 ELR25 4.7K ELR25	
C1 C2 C3 C4 C5 C6 C7 C8 C9 C10 C11 C12 C13 C14 C15 C16 C17 C18	ELECTROLY CERAMIC CONNECTOR	100 16V 220P 50V SL 220P 50V SL 220P 50V SL 15P 500V SL 220P 50V SL 220P 50V SL 0.0047 50V B 15P 500V SL 0.0047 50V B 0.001 500V B 0.5P 500V SL 2P 500V SL 33P 500V SL 37 500V SL 27P 500V SL	
P2 P3 P4	CONNECTOR CONNECTOR CONNECTOR	TMP-P01X-A1	
В1	P.C BOARD	B-503	

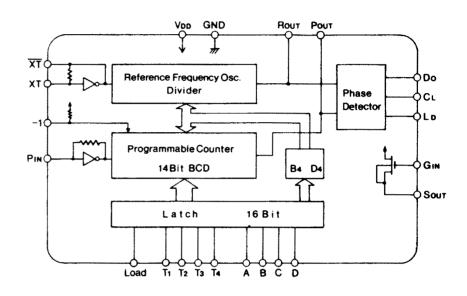
IC RATINGS

TC-9123P (FM/AM SYNTHESIZER TUNER PLL)

PIN CONNECTION

BLOCK DIAGRAM



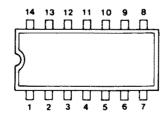


Maximum rating

Item	Symbol	Rating	Unit
Power supply voltage	Vaa	-0.3∼9.0	V
Input voltage	Vin	-0.3 ~ V _{DD} +0.3	V
Operation temperature range	Tope	−30 ·∼ +70	°C
Storage temperature range	Тѕтс	−55 ~ +125	°C

TC4001 (QUAD 2-INPUT POSITIVE NOR GATE)
TC4013 (DUAL D-TYPE FLIP-FLOP)
TC4028 (BCD TO DECIMAL DECODER)
TC4069 (HEX INVERTER)

PIN CONNECTION



Maximum rating

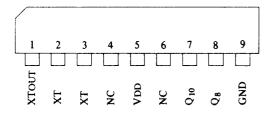
ltem	Symbol	Rating	Unit
Power supply voltage	σαV	Vss −0.5 ~ Vss +20	٧
Input voltage	Vin	Vss −0.5 ~ Vpp +0.5	٧
Output voltage	Vout	Vss −0.5 ~ Vpp +0.5	٧
Input current	lin	±10	mΑ
Permissible Dissipation	Рь	300	mW
Storage temperature range	Тѕтс	−65 ~ 150	°C
Lead temperature/time	Tsol	260°C · 10Sec.	- 140

TC-5082 (OSCILLATOR AND 10-STAGE DIVIDER)

Maximum rating

ltem	Symbol	Rating	Unit
Power supply voltage	VDD	10 `	V
Input voltage	Vin	-0.3 ~ V _{DD} +0.3	V
Operation temperature range	Topr	−30 ~ 75	°C
Storage temperature range	Tstr	−55 ~ 125	°c

PIN CONNECTION

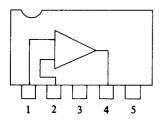


BA401 (FM/IF LIMITER)

Maximum rating

Item	Symbol	Rating	Unit
Power supply voltage	Vcc	15	٧
Output voltage	Vouт	24	V
Input voltage	Vin	±3	٧
Operation temperature range	Topr	−25 ~ +75	°c
Storage temperature range	Тѕтс	−55 ~ +125	°C

BLOCK DIAGRAM

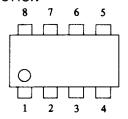


MJM4558D (DUAL LOW NOISE AMP.)

Maximum rating

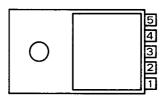
Item	Symbol	Rating	Unit
Power supply voltage	VaaV	18	V
Input voltage	Vin	15	V
Operation temperature range	Торт	−20 ~ +75	°C
Storage temperature range	Тѕтс	-40 ∼ +125	°c

PIN CONNECTION



μ PC2002 (5.4-W AUDIO POWER AMP.)

PIN CONNECTION

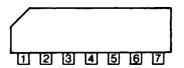


Maximum rating

ltem	Symbol	Rating	Unit
Power supply voltage (surge)	Vcc 1 (50ms)	40	V
Power supply voltage (no-load)	Vcc 2	28	V
Power supply voltage (operation)	Vcc 3	18	V
Circuitry current (continuous pulse)	Icc(PEAK)1	3.5	Α
Circuitry current (single pulse)	Icc (PEAK)2	4.5	А
Package Dissipation	Po(Tc=90°C)	75	W
Operation temperature	Topr	−30 ~ +75	°c
Terminal-terminal voltage	Тѕтс	−40 ~ +150	°c

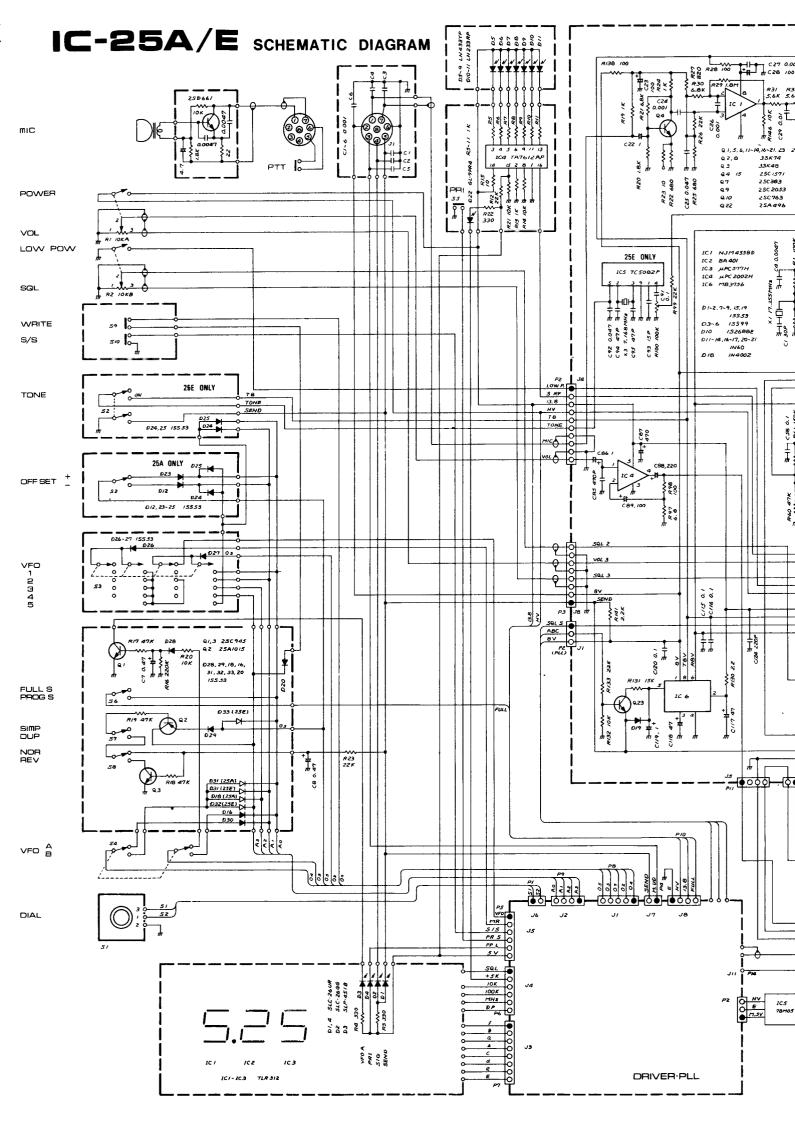
μ PC577H (FM-IF AMPLIFIER)

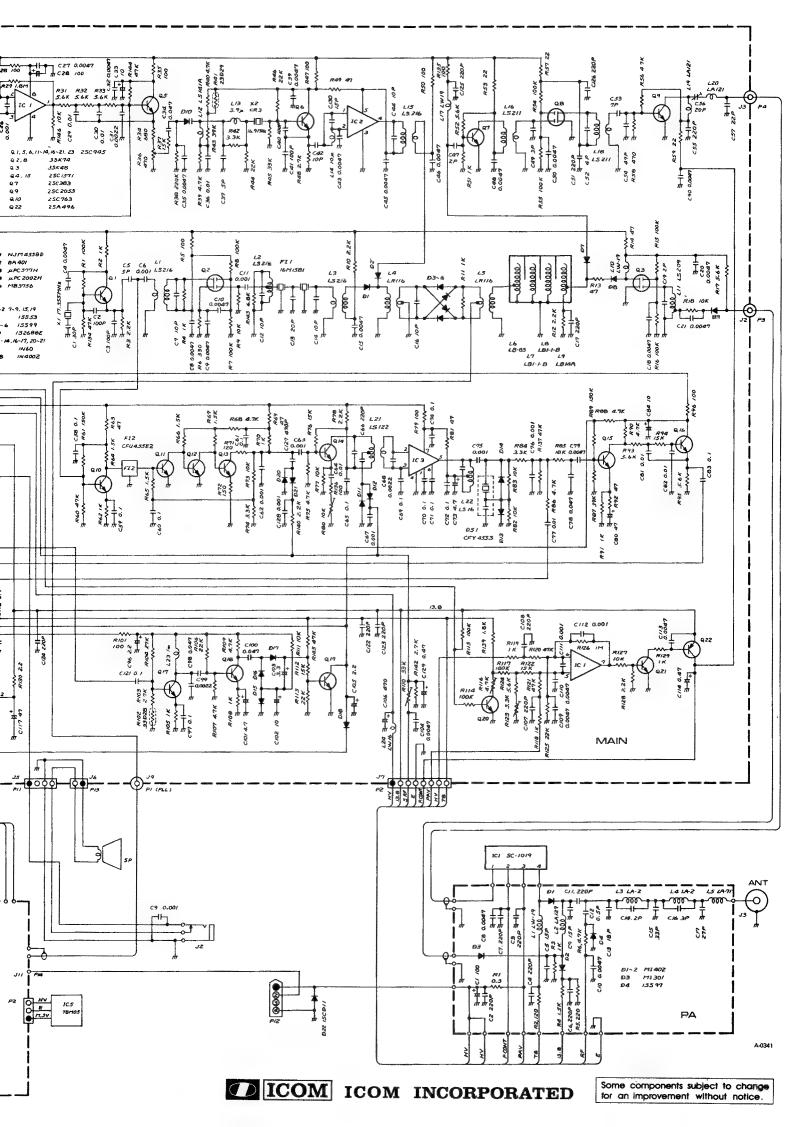
PIN CONNECTION

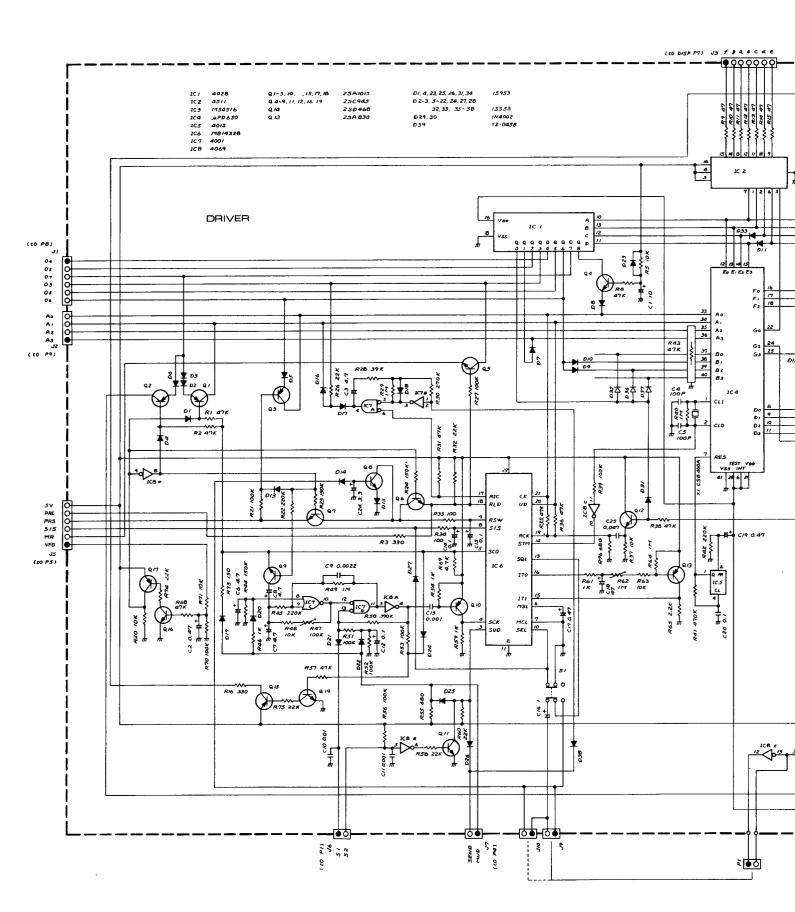


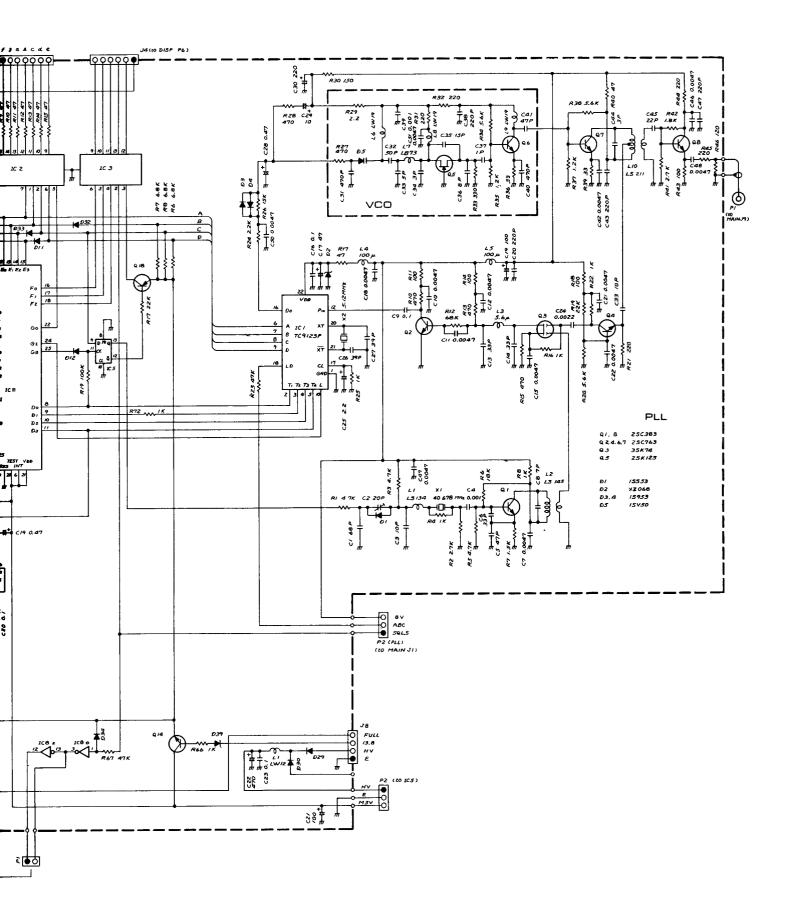
Maximum rating

Item	Symbol	Rating	Unit
Power supply voltage	Vcc	15	V
Terminal-terminal voltage	Vin	±3.0	V
Permissible Dissipation	Рь	300	mW
Operation temperature	Торт	−20 ~ +75	°c
Storage temperature	Тѕтс	−40 ~ +125	°C











MIC BOARD

SWITCH BOARD (1) 在 # 83 25 (ENB) 2M B-2 W S X 88 TIA 44 D28 (2) BOARD

DISPLAY BOARD (1)

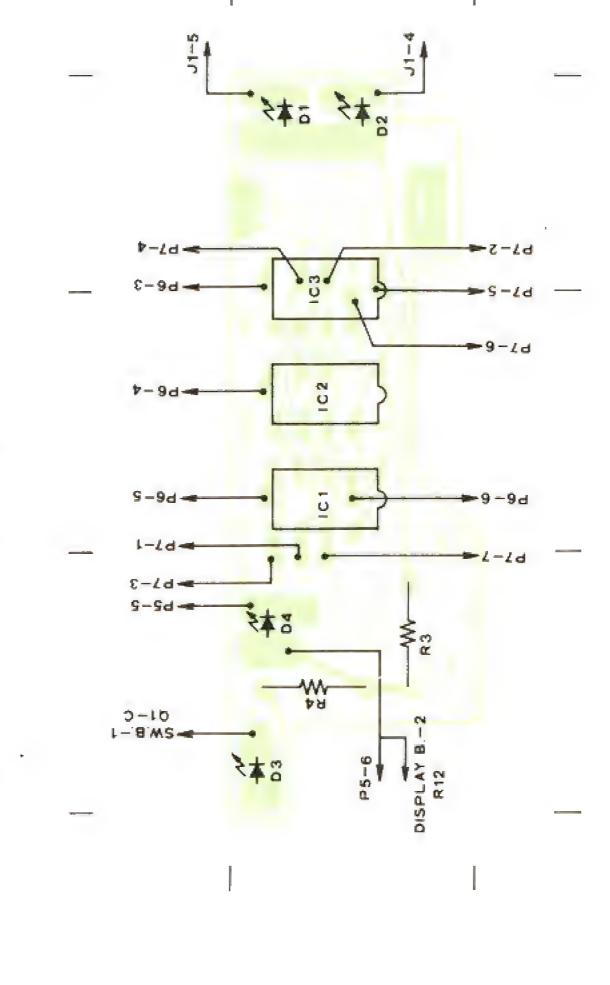
Z-01d-

(AU3)SEQ 150 810 (A2U)

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-SW. B.-

910

-SW.B-

55-2 25-C(N2V) 2M B-2

1-6d

29 MS

D3 - K

P2-1 (USA)

SW.B.-1

000 ECB

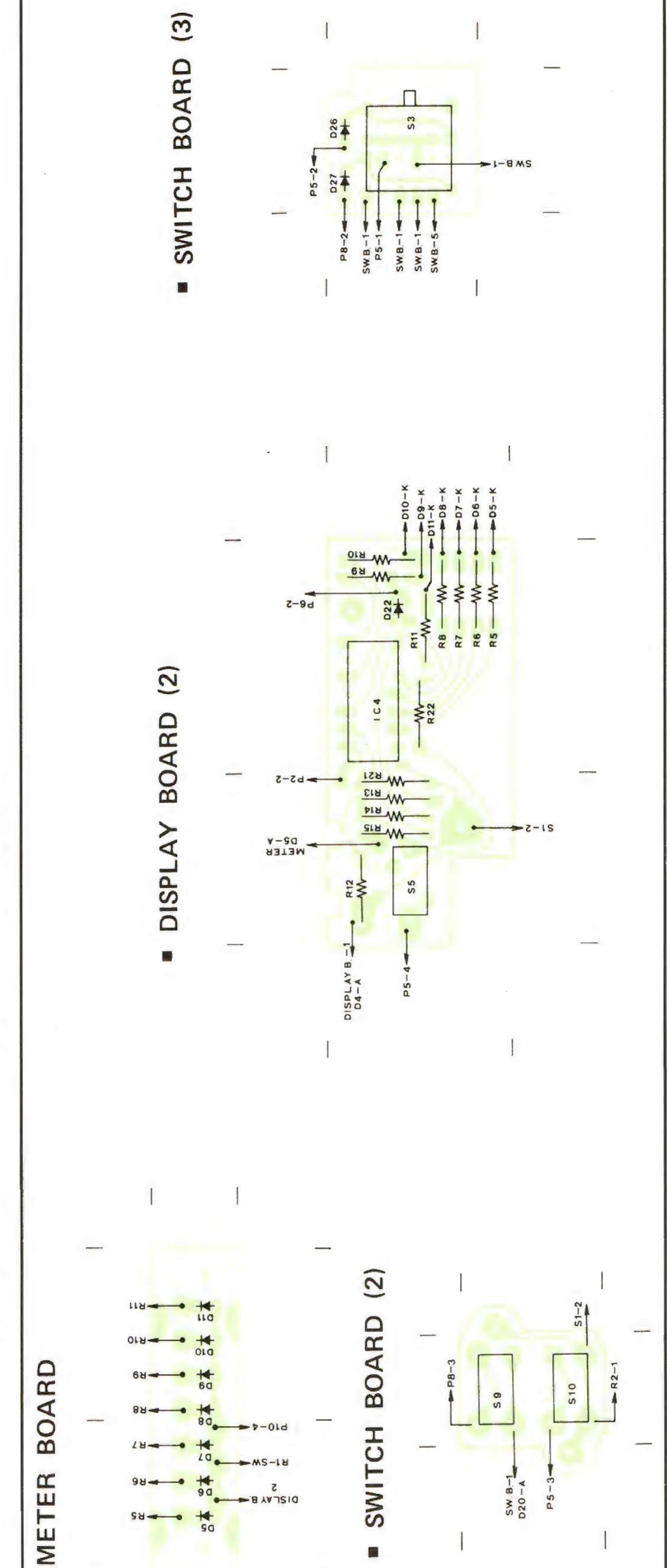
02

EC B

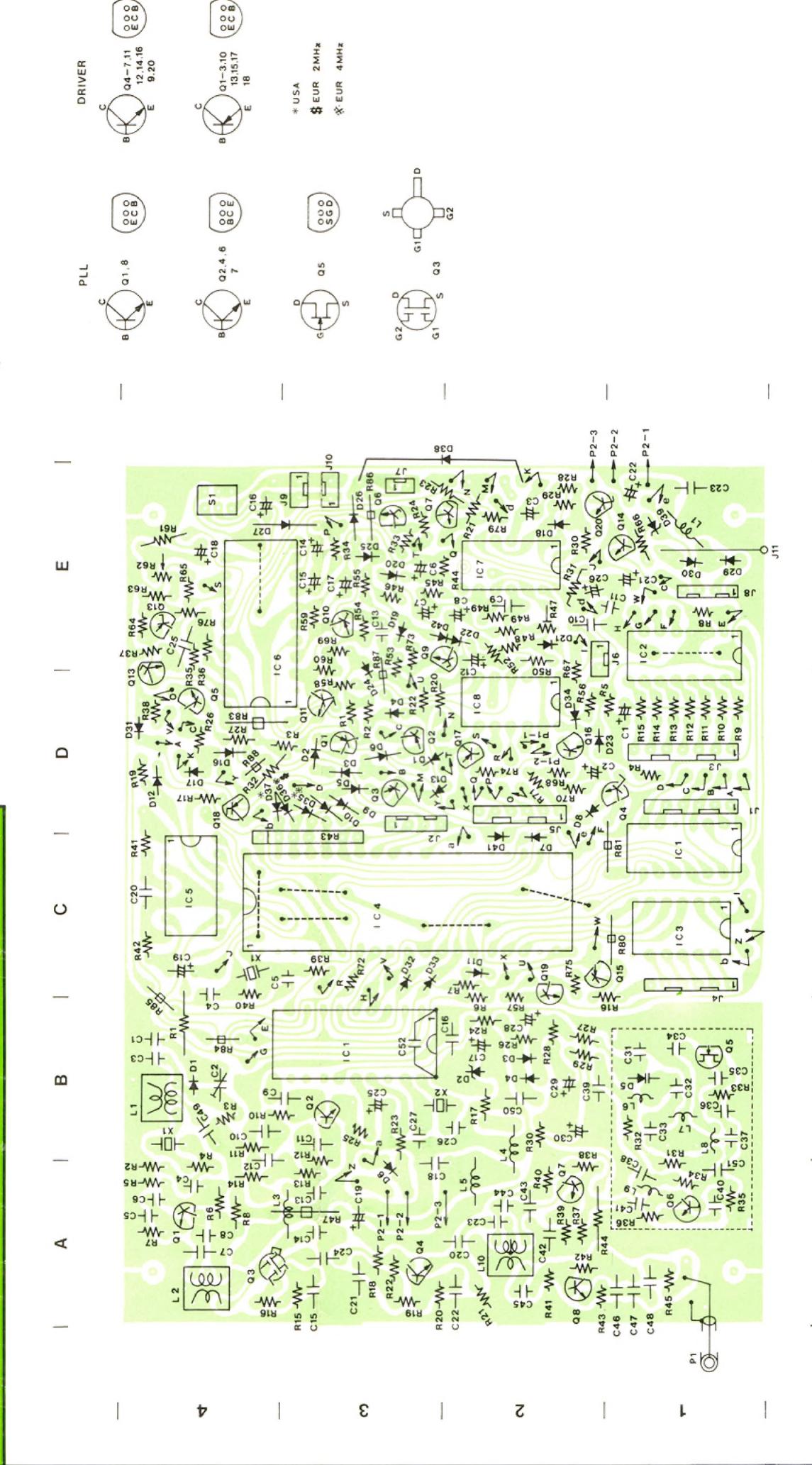
(EUR) MIC B

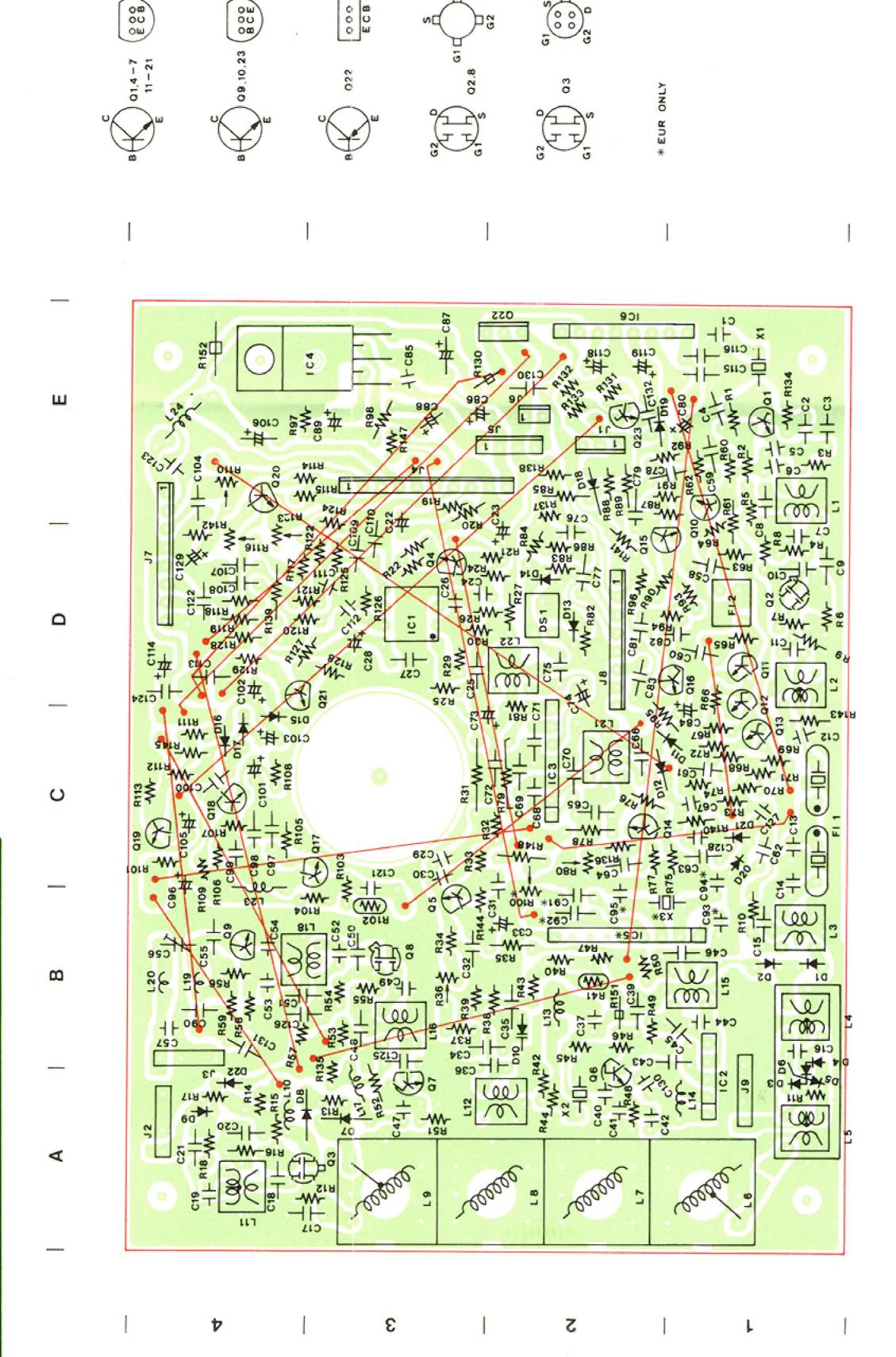
029-A (USA) P2-5 (RUR)

BOARD LAYOUT

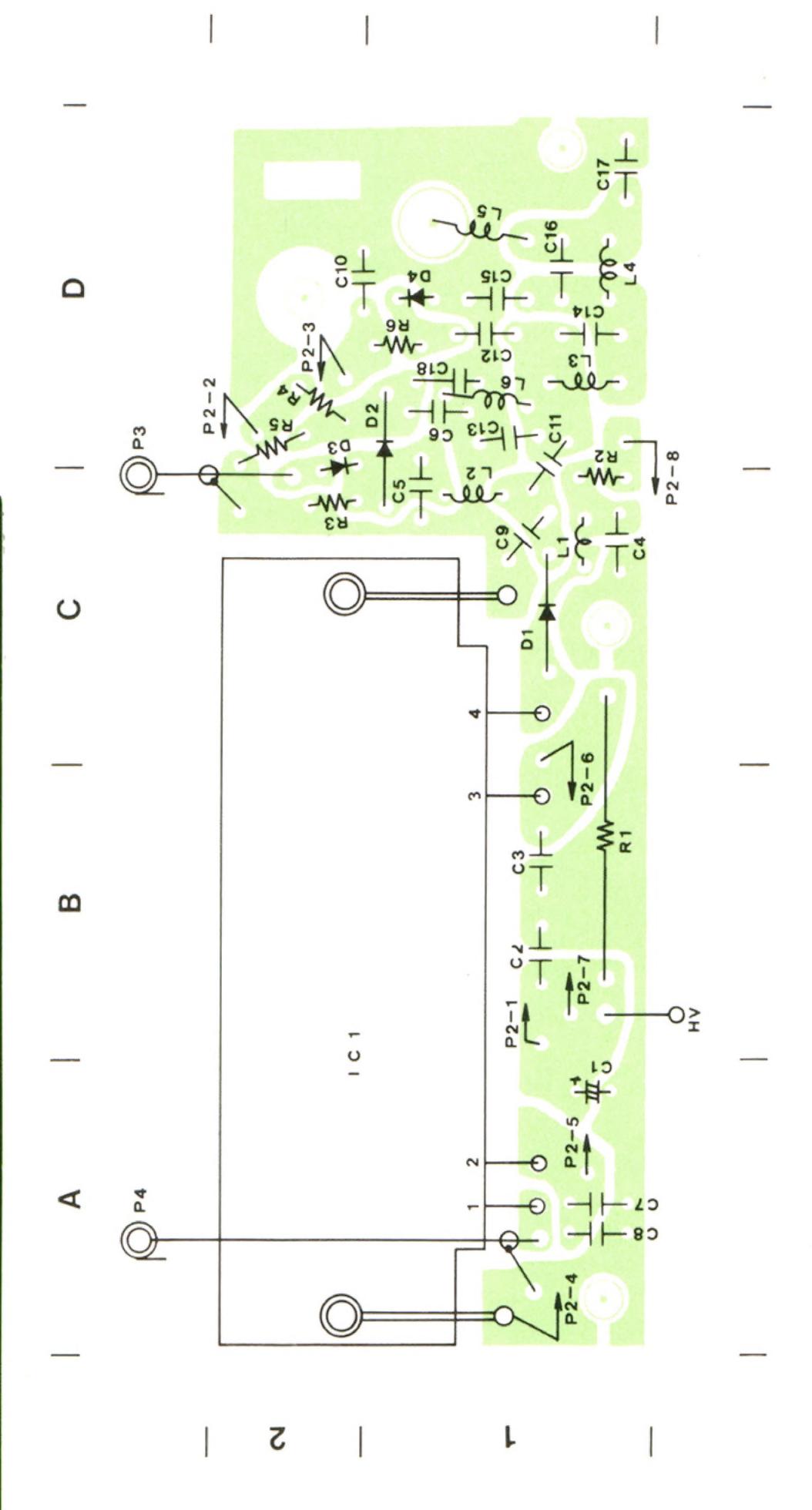


PLL . DRIVER UNIT





OOO ECB



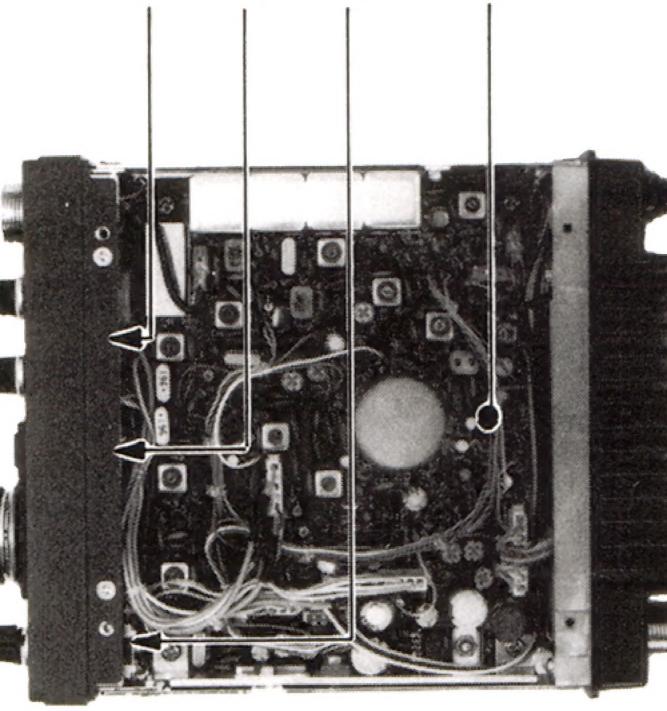
UNIT LAYOUT

SWITCH BOARD

METER BOARD

DISPLAY BOARD 1.2

PLL. DRIVER UNIT



SWITCH BOARD

SWITCH BOARD

MIC BOARD

SWITCH BOARD

MAIN UNIT

PA UNIT -